INDUCTION SYSTEM FOR A PROTON BUNCH ACCELERATION IN SYNCHROTRON *

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Abstract

Recently the first induction acceleration of a single proton bunch of 2.3E10ppp was demonstrated [1], where it was accelerated from 500MeV to 8GeV in 1.9sec in KEK-PS. This paper describes the induction system for a synchrotron and a beam confinement experiment as a second step of the proof-of principle (PoP) experiments of a "induction synchrotron" [2], where a proton beam with a bunch length of 500ns was longitudinally confined by the induction barriers for 500msec, is introduced..

INDUCTION SYSTEM FOR SYNCHROTRON

The induction system required for the proton bunch acceleration was integrated in September 2004. The installed components are: 1) the switching power supplies [3], 2) the induction cells [4], 3) the trigger system, and 4) the interlock system. The entire system is described in an equivalent circuit, shown Figure 1.

Switching Power Supply

A DC power supply, which is connected to the switching power supply (SPS) via a power cable of 100m, transforms the input voltage of AC 400V to a maximum output voltage of 3.0kV by switching IGBT.

The switching power supply is formed from a fullbridge switching circuit. Each switching arm consists of 7 series of MOS-FET, so as to withstand a maximum voltage of 2.5kV with an arm current of 18A. A rise-time and fall-time of the voltage pulse is 30ns each. A maximum repetition rate of the SPS is limited by a temperature rise of the MOS-FETs due to switching loss, which is estimated to be 150W/3.7cm². An indirect watercooling copper block is contacted to a surface of the MOS-FETs. A total heat deposit of 4.2kW is removed by a water flow of 12L/min. Three SPSs, shown in Figure 2, are placed 40m away from the accelerator in order to avoid radiation damages on MOS-FETs. The pulse power is transmitted to the induction cell through the 40m transmission cable of 1250hm.

Gate signals of the MOS-FETs are transferred from the trigger system, which is described below, in the form of optical signal. The gate terminal of the each MOS-FET is insulated from a ground. An electric power, which is required for driving the gate of the MOS-FET, is supplied from DC-DC converters. An input terminal of the SPS is a group of film condensers to keep the output voltage from a perturbing change in the charged voltage.

Induction Cell

One induction cell consists of 6 magnetic cores, which are made of nanocrystalline alloy (FINEMET). Each magnetic core has an outer diameter of 500mm, inner diameter of 225mm, and thickness of 15mm. A diameter of a induction cell beam pipe is 150mm. A ceramic gap of 5cm in length and 8mm in thickness is employed. Silicone oil, which is helpful to decrease a stray capacitance, is a coolant for the induction cell. A flowrate of the coolant of the induction cell is 70L/min, providing a maximum cooling capacity of 20kW. The induction cells aligned on the beam line are shown in Figure 2.

An equivalent electric circuit of the induction cell is represented as a parallel connection of an inductance of 110μ H, a capacitance of 280pF, and a resistance of 3000hm. These numbers have been obtained from a fitting result of a measurement of a beam-induced voltage of the induction cell [4].

A matching resistor of 2100hm, which is determined so as to improve the impedance matching to the transmission cable around the revolution frequency of 667kHz– 882kHz, is connected in parallel to the induction cell. A peak impedance of the (induction cell + matching resistor) is 1410hm at 5.9MHz, and an obtained quality factor of the load is 0.1. The matching load is indirectly water-cooled. A flow-rate of the cooling water is 5L/min, providing a maximum cooling capacity of 15kW. The single induction cell is capable of generating an acceleration voltage of 2kV. Three cells have been installed for the first POP experiment.



Figure 1: Signal Flow Diagram of Induction System



Figure 2: Switching Power Supplies (left) and Induction Cells with Matching Load (right)

Trigger System

The trigger system consists of four electric components: 1) voltage comparator, 2) delay module, 3) pulse-pattern generator, and 4) electric/optical signal converter. Signal flows in the trigger system are shown in Figure 1. The wall current monitor (WCM) generates an initial trigger signal. A rising edge of the WCM signal is detected by the voltage comparator in the accelerator central control room (CCR), and a timing signal of rectangular shape are generated. Subsequently the timing signal is transferred to the delay module, in order to keep the generation of the induction voltage synchronized with the beam passing through the induction cell. Then, the delayed signal is transferred to the pulse-pattern generator. A pulse width of the induction voltages can be varied by changing on/off timing of the switching arm in the SPS. A delayed negative voltage to a positive induction voltage can be controlled in a similar way. Basically the pulse width of the positive / negative voltage must be same, in order to avoid input of an equivalent DC voltage.

The pulse pattern signal is transferred to the electric/optical signal converter placed near the SPS through the 80m-log transmission cables of 500hm. The converter and the SPS are connected by the 10m-long optical fiber cable. These optical signals serve to drive MOS-FETs as their gate signals.

Interlock System

Each component of the induction system is endowed with an interlock system to protect itself from any malfunction. A programmable logic controller (PLC) is used to manage interlock signals. The interlock system detects fatal troubles, such as a temperature rise in the induction cell, insufficient oil flow, insufficiency water flow in the SPS/the matching resistor, and overvoltage/over-current in the MOSFETs.

In a case of safety, an input of an enable signal from the interlock system is required in order to output trigger signal. And a manual input of a reset pulse to the interlock system is required in order to resume from the trouble.

BEAM-RF PHASE MEASUREMENT

A measurement of the bunch position in the rf-phase, which is monitored by the phase detector integrated in the rf-feedback control system through the entire stage of acceleration, is indispensable to experimentally confirm the induction acceleration of a proton bunch in the hybrid system, where the induction acceleration system is combined with the rf system. When the beam acceleration is ideal, the bunch should stay on a specific phase of an rf, $\phi_s = \sin^{-1}[\rho C_0(dB/dt)/V_{rf}]$, where ρ , C_0 , B, and V_{rf} are the bending radius, circumference of an accelerator ring, bending field, and amplitude of the rf, respectively. An actually obtained output-signal of the phase detector V_{out} , which is closely related to ϕ_s , is affected by a finite gain factor and off-set. Therefore a calibration of the phase detector has been carried out by the practical method of varying the rf amplitude and observing the output-signal of the phase detector.

Assuming a temporal evolution of a linear gain factor a(t) and off-set term b(t), $V_{out}(t,\varphi_s(V_{rf}))$ is written by $V_{out}(t,\phi_s(V_{rf})) = a(t)\phi_s(V_{rf}) + b(t)$. By varying V_{rf} , data array of the theoretically estimated $\varphi_s(V_{rf})$ and V_{out} at a certain instant t of the acceleration period can be obtained as is shown in Figure 3. In this figure t = 500ms for an example. Finally a(t) and b(t) is obtained by applying least-squares method to these plots at each instant, which is shown in Figure 4.



Figure 3: An Example of $\Delta \phi$ signal at t = 500 msec



Figure 4: Temporal evolution of the linear gain factor a(t)and offset term b(t)

Theoretically estimated accelerating phases for different Vrf are given in Table 1. And the temporal evolution of a calibrated output signal of the phase monitor by rf acceleration is shown in Figure 5. A comparison of the required accelerating phase and the calibrated output signal indicates that the output signal of the phase detector is valid for measuring the acceleration phase.

Tabla 1	Estimated	Acceleration	Dhaca
I able i	Estimated /	Acceleration	Phase



BEAM CONFINEMENT BY INDUCTION VOLTAGE

The beam confinement using the induction step barriers is crucial to achieve a super-bunch for future applications [5]. A 5.0E10 ppp of proton beam was successfully confined in a shallow notch potential generated by the induction voltage $V_{barrier}$ of 4.7kV, with pulse width T_1 of 400ns. An outline of the experiment is shown in Figure 6.

Assuming a rectangular barrier voltage, the bucket height in energy ΔE_{max} is expressed in terms of the pulse voltage $V_{barrier}$, a pulse width T_l , a revolution period T_0 , a velocity of the reference particle β , energy of the synchronous particle E_0 , and a slippage factor η ,

$$\Delta E_{\text{max}} = \sqrt{\frac{eV_{barrier}T_1}{T_0} \frac{2\beta^2 E_0}{|\eta|}} \cdot$$

This experiment was carried out at the injection energy (500MeV) of the KEK-PS, where $T_0 = 1.51 \mu s$, $\beta = 0.751$, $E_0 = (500+938)$ MeV, and $\Delta E_{max} = 0.18\%$. An experimental

result is shown in Figure 7, where a plotted line from bottom to top in the graph represents a time evolution of the beam shape in a longitudinal direction. In this experiment, 1.0E11ppp of proton beam with maximum energy spread of 0.15% is injected. From a comparison of the energy spread of the beam and the depth of barrier potential, all injected particle was to be confined within the barrier voltage. However, 50% of injected beam was lost during 460ms. Possible causes of the beam loss are under investigation.



Figure 6: Beam Confinement by Induction Voltage



Figure 7: Bunch Shape during Experiment of Beam confinement

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