



LHC Injectors Upgrade

Low Level RF Control Algorithms for the CERN LINAC4

P. Baudrenghien, CERN SY-RF

R. Borner, B. Bielawski

The LLRF design has started in 2009. Several persons have contributed. In order of appearance in the project: *A.K. Bhattacharrya, D. Stellfeld, J. Noirjean, G. Hagmann, D. Valuch, J. Galindo, M. Ojeda Sandonis, V. Costa, B. Kremel, J. Simonin, K. Adrianek, M. Andersen*

Thanks to *M. Crofford, L. Doolittle* and *T. Hardek*, from SNS for the very interesting visit in 2009

Thanks to *K. Fung* from TRIUMF for discussions on the AFF during LLRF17 workshop.



Content



- Linac 4
- Field regulation. How precise must we be ?
- Field perturbations. What are the causes?
- Linac4 LLRF architecture
- Classic RF feedback (2013-2018)
- Linear Quadratic Gaussian (LQG) regulator (2019-)
- Adaptive Feed Forward (AFF) (2019-)
- Klystron Polar Loop (KPL) (2020-)
- Conclusions





Linac 4 [1]

- H- ions
- Acceleration to 160 MeV
- 352.2 MHz RF
- Injection of protons into the PSB synchrotron, after stripping foil
- 1.2 s repetition time
- At each pulse it can accelerate up to 600 μ s of beam consisting of four batches (one per PSB ring) spaced by 2 μ s
- It includes a chopper at 3 MeV to
 - remove the bunches that would fall outside the 1 MHz PSB bucket ($h=1$)
 - create empty 2 μ s long beam gaps to cope with the switching time of the distributing magnet that routes the Linac beam to the four superposed PSB rings
- As a consequence, the cavities see *strong transient beam loading* as the beam intensity changes from zero to maximum beam current (presently 25 mA DC) in just 3 ns.





Field stability margins

- Beam dynamics simulations have been carried out, early in the machine design, to define the maximum level of acceptable RF phase and amplitude jitter
- Table reproduced from [2]

	Amplitude Jitter (Uniform)	Phase jitter (Uniform)	Static field error – (Uniform)	Field unbalance for coupled cavities
RFQ	$\pm 1\%$	n/a		n/a
MEBT bunchers	$\pm 1\%$	$\pm 1 \text{ deg}$	n/a	n/a
DTL	$\pm 0.5\%$	$\pm 0.5 \text{ deg}$	$\pm 2\%$ (random gap field error)	n/a
CCDTL	$\pm 0.5\%$	$\pm 0.5 \text{ deg}$	$\pm 2\%$ (tilt over 3 modules)	$\pm 5\%$
PIMS	$\pm 0.5\%$	$\pm 0.5 \text{ deg}$	$\pm 5\%$ (tilt)	Not investigated
Transfer line debuncher	$\pm 1\%$	$\pm 1 \text{ deg}$	n/a	n/a

- So, the LLRF had to implement field regulation within **1 RF degree, 1% amplitude pk-pk** during beam pulse, for the nominal **40 mA DC** beam intensity.



Field perturbations (1/2)

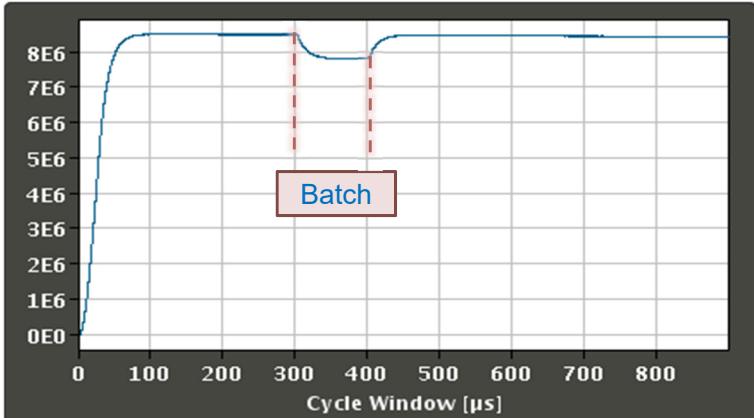
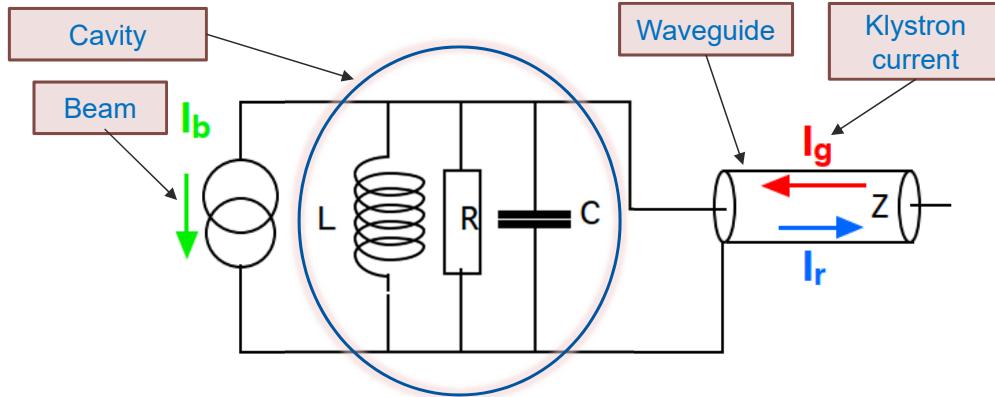
- Environmental causes
 - Temperature drifts will expand the tunnel floor and change cavity spacing
 - Temperature drift will affect **cable length** (and phase shift)
- Cavities
 - Will be subject to vibration and microphonics that may fall outside the range of the tuners
- Power amplifier
 - Solid state amplifiers show **gain/phase ripples** caused by the noise in their DC supply
 - Even worse with klystrons. LEP tubes show 0.1 dB and 8 RF deg/ % HV voltage drift. A **5% change in HV** results in **6 % amplitude** and **40 RF deg change** in cavity field (voltage)
 - The slow drift during the pulse is called *klystron droop*
 - The higher frequencies come from noise in the generation of HV (rectifiers).





Field perturbations (2/2)

- Beam loading
 - A cavity is a resonant circuit excited by **two currents** [3]: The RF amplifier I_g (for generator) and the beam I_b
 - This is **by far the largest perturbation** in the L4 cavities
 - The **25 mA DC** beam current induces almost **1 MV** in the CCDTL1 cavity. Shown is cavity voltage amplitude without regulation. For nominal **40 mA** the beam induced voltage will be 1.6 MV, that is **20 %** amplitude variation. To respect the specifications **the beam loading peak must be reduced by 20 linear minimum.**

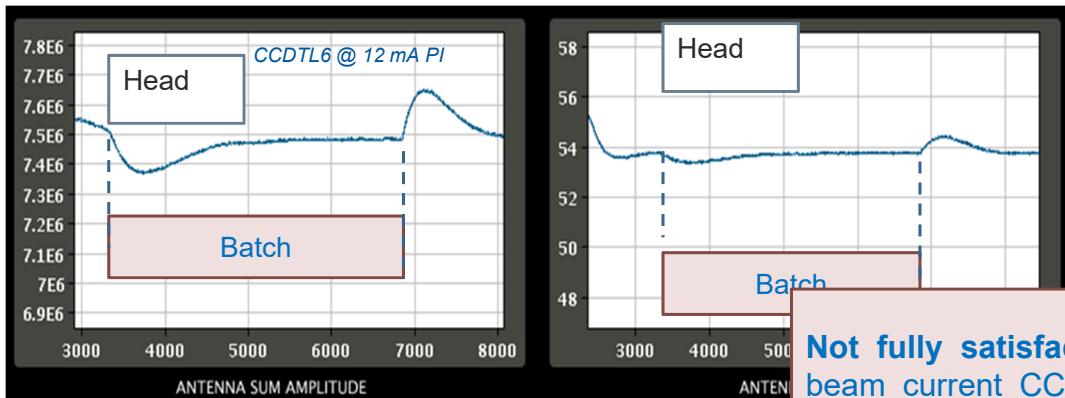


CCDTL1 voltage amplitude (in V) without regulation with a 25 mA DC beam current.



Classic RF feedback 2013-2018

- The LLRF was first commissioned with a classic RF feedback [5]
- The LLRF consisted of a *PI Controller*, including a proportional (P) and an integral (I) gains
- The limitation of this simple regulator comes from the loop delay including TX, circulator, waveguide, antenna cable and LLRF latency. **Total around 2.5 us.**
- The plots show the performances with PI controller and 12 mA beam current (2017).



Antenna Sum amplitude and phase sampled at Frf/16 (45.45 ns/sample).

The regulation is slow: the transient affects the first 40 μ s of beam.

For CCDTL6 we have 0.1 MV pkpk (1.3%) fluctuation in voltage and 0.5 deg in phase .

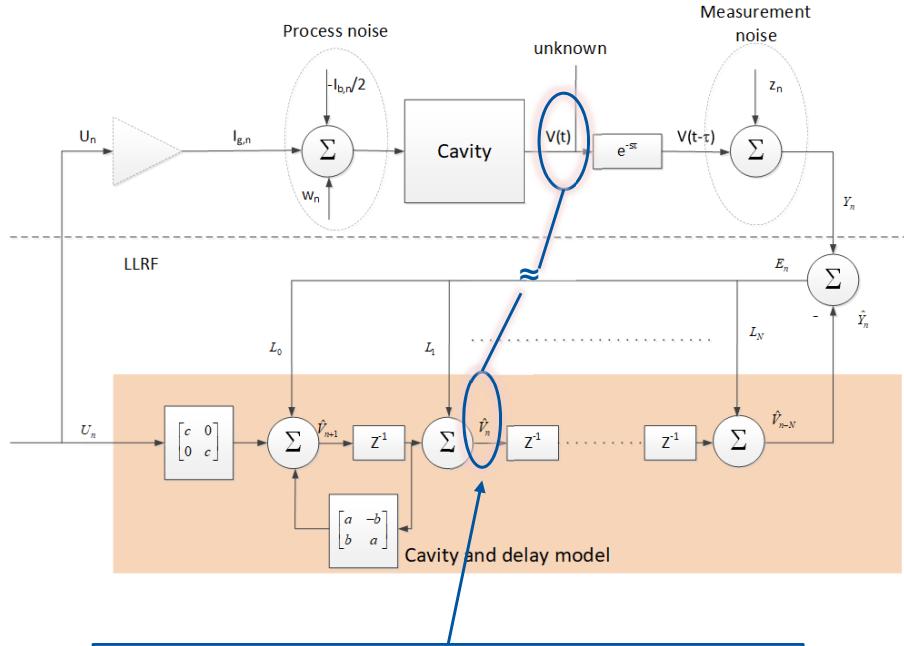
Not fully satisfactory: Scaling to nominal 40 mA beam current CCDTL6 would exceed tolerance by factor 4 in amplitude and almost 2 in phase.



Model based. LQG regulator [6]. 2019

- **Step 1. The Kalman predictor**

- Implement in the LLRF a **model of the cavity plus delay**
- Feed it with the **same drive** as the cavity amplifier
- Of course the model output will not be exactly the cavity output. The difference comes from inexact model plus the **noise injected into the cavity and absent in the model**. This is called **Process noise**. In our case that is mainly the **beam loading**
- **Update** the model states by **comparing** the model and cavity output
- There will be noise at the cavity output, the so-called **Measurement noise** (small in our case)
- The optimum matrix L depends on the relative importance of process noise and measurement noise. If process noise is high compared to measurement noise (our case), the **correction will be applied strongly as we have much confidence in the measurements**.



What have we gained?

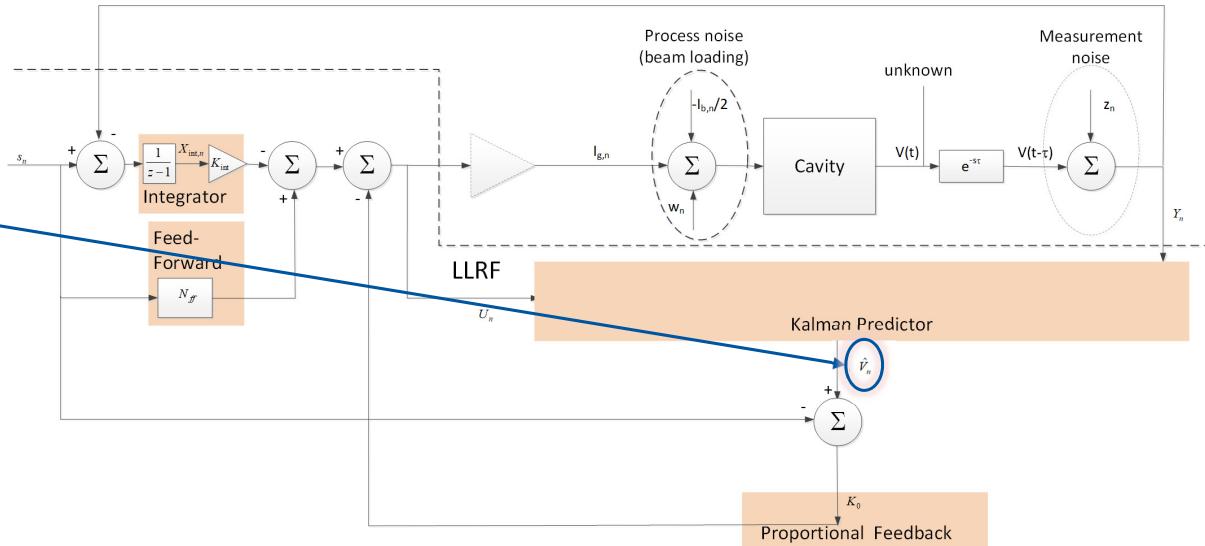
We have reconstructed a signal in the LLRF that is a good **estimate** of the cavity voltage and ...**available without delay**





• Step 2. The Feedback

- We now implement a proportional-integral (PI) feedback using the cavity voltage **estimate**
- We optimize the feedback using the Linear Quadratic Regulator method (LQR).



What have we gained?

As long as our **estimate** is very close to the cavity voltage, we have a feedback **without delay**.

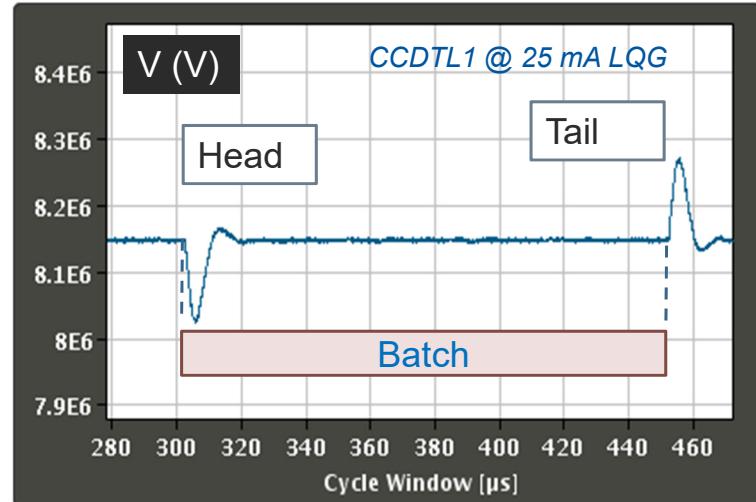
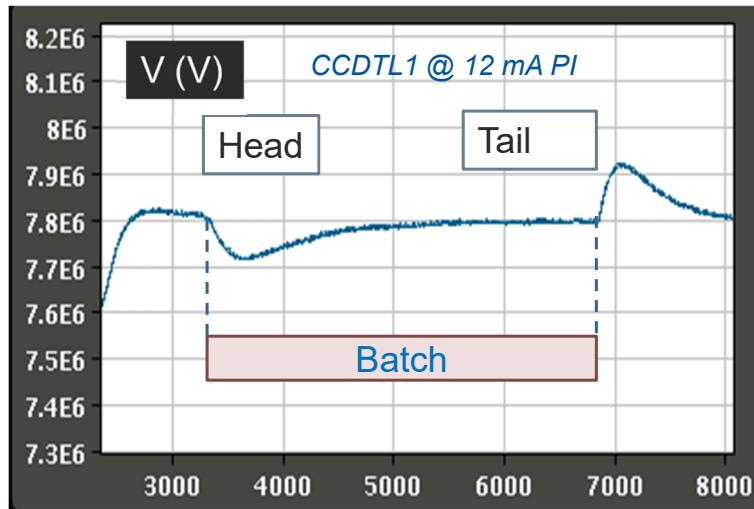
The Kalman Predictor must react significantly faster than the LQR.

In our optimization we have

- Kalman Predictor responding in 200 ns
- LQR responding in 900 ns.



LQG feedback installed begin 2019



Voltage (Antenna Sum)

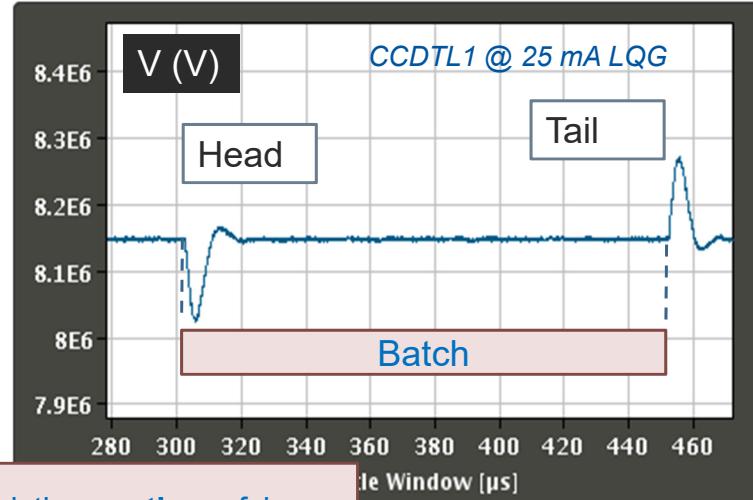
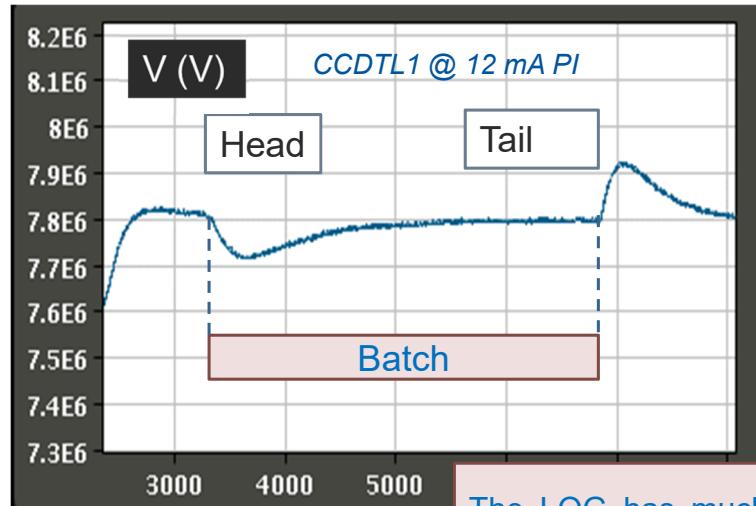
Left: performance measured with PI controller (2017), 12 mA DC (45.42 ns/sample)

Right: idem measured with LQG regulation (2019), 25 mA DC

The pk-pk amplitude is reduced by almost two (accounting for the different beam intensity) and **the regulation is much faster**: Only the first 5 μ s of beam are now affected.



LQG feedback installed begin 2019



The LQG has much reduced the portion of beam affected...but we still observe a significant but short transient at the beginning of the batch.

Voltage (Antenna Sum)

Left: performance measured with

Right: idem measured with LQG regulation (2019), 25 mA DC

The pk-pk amplitude is reduced by almost two (accounting for the different beam intensity) and the regulation is much faster: Only the first 5 μ s of beam are now affected.



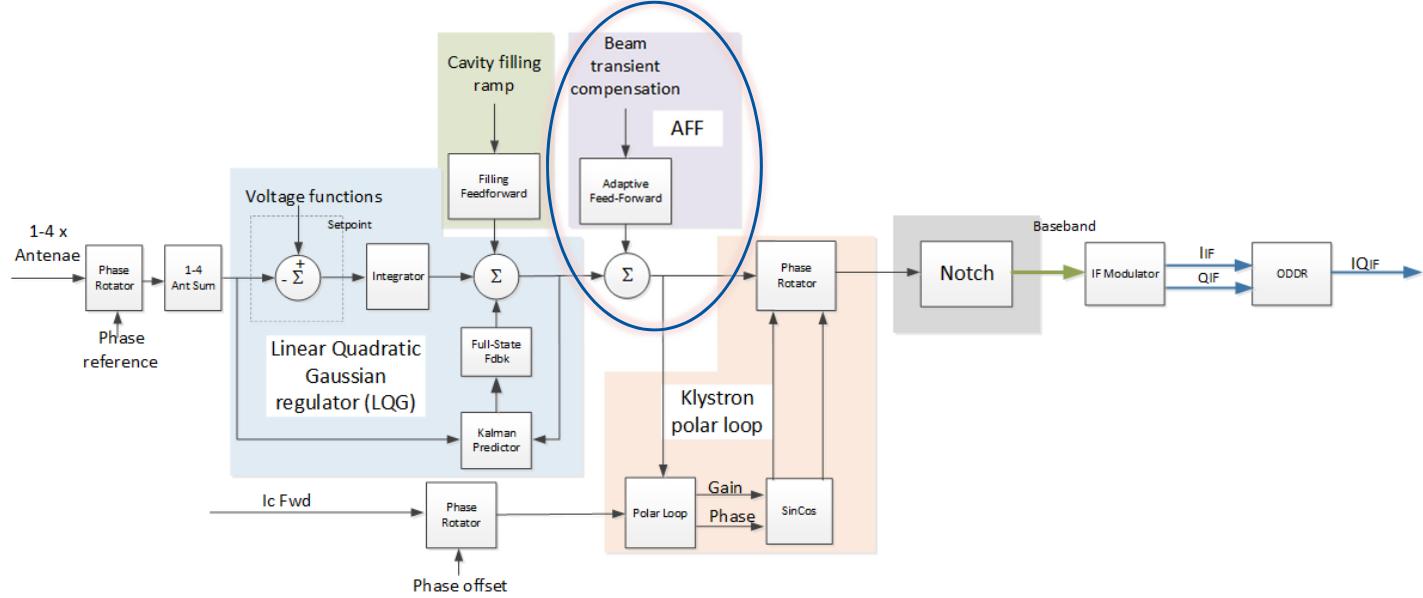
Adaptive Feed Forward (AFF). 2019

- As long as we react pulse per pulse, we cannot completely cancel the transient at the head of the batch due to causality. We cannot **anticipate** the required increase of klystron power
- Assuming that this transient is similar from pulse to pulse, we could base our correction on the observation of previous pulses and...**anticipate** for the present pulse
- That is the idea behind **Adaptive FeedForward (AFF)**. It works great for **repetitive** perturbations
- What perturbations are pseudo-periodic (similar from pulse to pulse)?
 - **Beam loading?** Yes, because source current does not change much from pulse to pulse
 - **RF noise caused by klystron HV ripples?** Not in the L4 as the modulators are not synchronized to the pulsing
 - **Vibrations** (microphonics)? No
- In L4 AFF is **very effective** for the compensation of the beam loading transient **at the batch head**.





AFF. How?

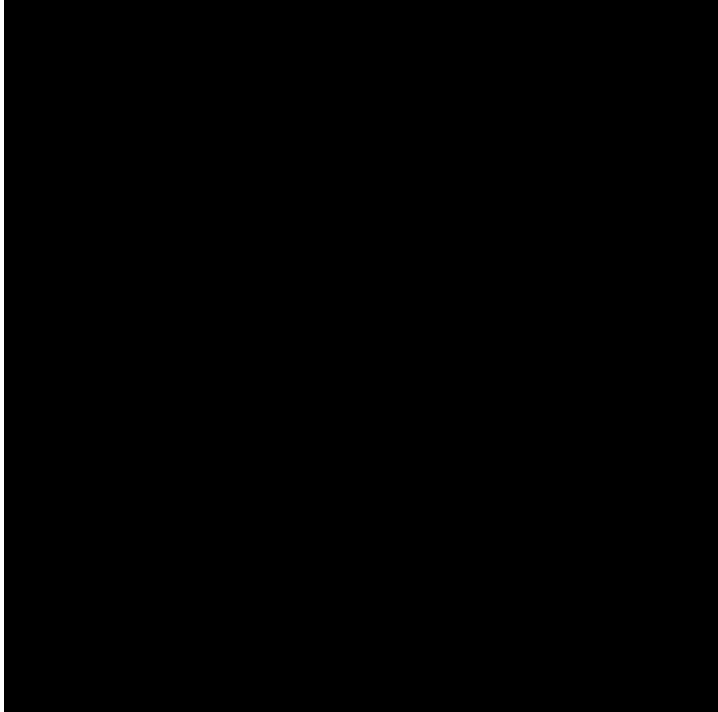


- The AFF is a correction based on the observation of the voltage error over the few previous pulses
- It is added to the klystron drive
- It is computed **off-line** (FEC). We use an algorithm similar to the one developed at TRIUMF [8]: We filter the voltage error with an impulse response **advanced correctly in the pulse**. Critical!
- It is loaded in the firmware at each pulse.

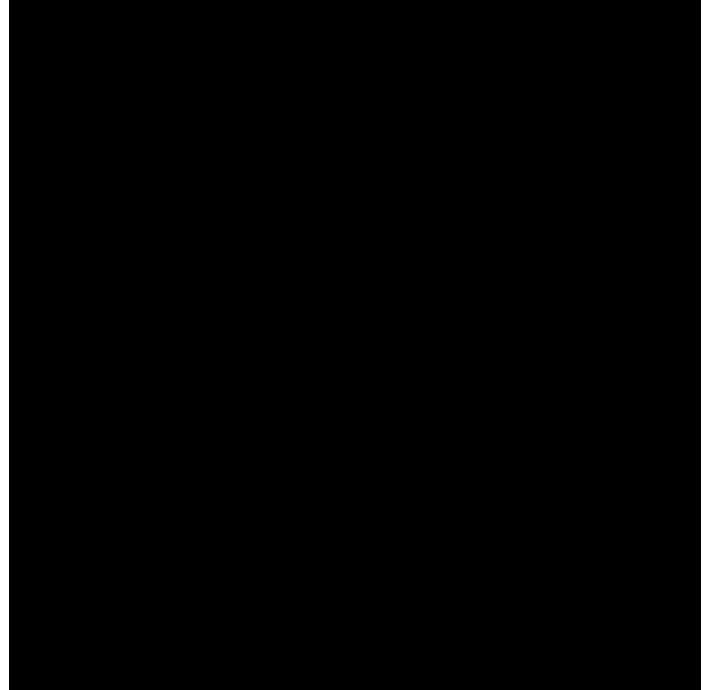


AFF Learning

Error $V_{\text{cav}} - V_{\text{set}}$
(I,Q)



Cavity voltage (V)



AFF correction
drive (I,Q)

Cavity phase (deg)

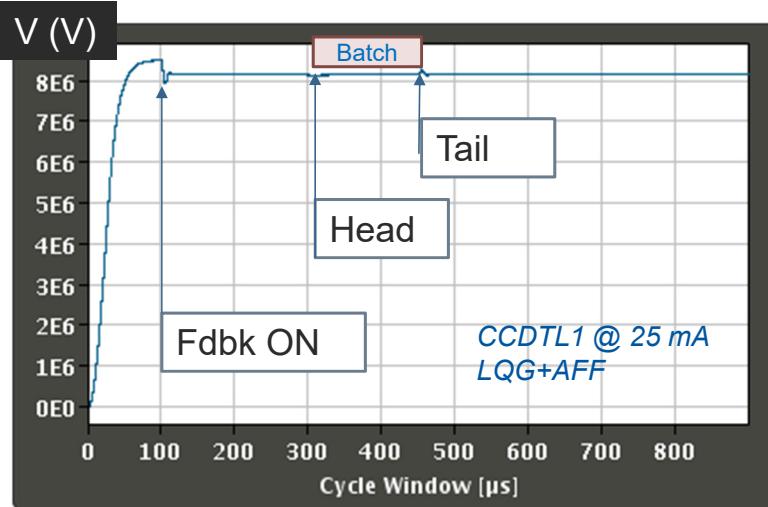
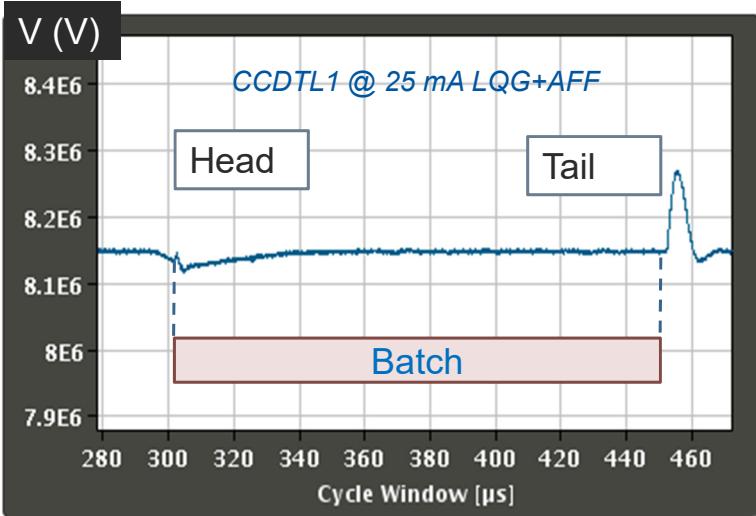
- I/Q error and correction (left) and cavity voltage amplitude/phase (right) during AFF learning
- Four batches, 100 us long, with 2 us gaps
- The correction is reset at beginning and in the middle of the record.

Linac'22





Overall performances

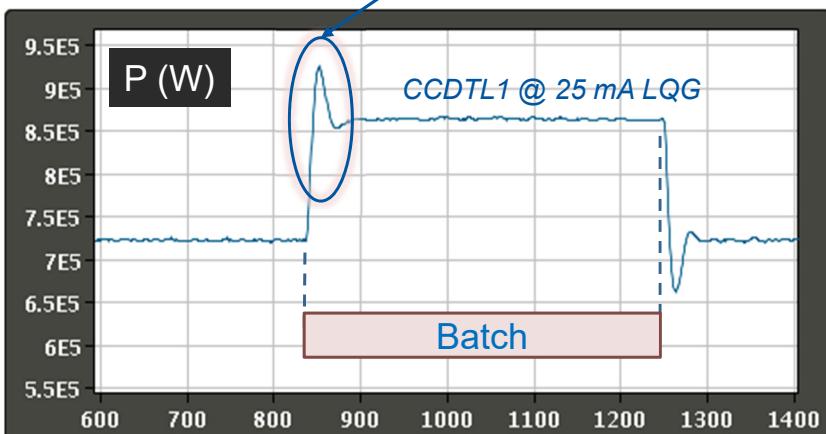


- Uncompensated beam loading 30 kV pk-pk, that is 0.375 % for 25 mA DC \rightarrow **0.6 % for nominal 40 mA DC current. Factor 2 better than specs....**
- The beam becomes barely visible in the overall pulse
- The AFF is switched off at the end of the batch, therefore the transient.

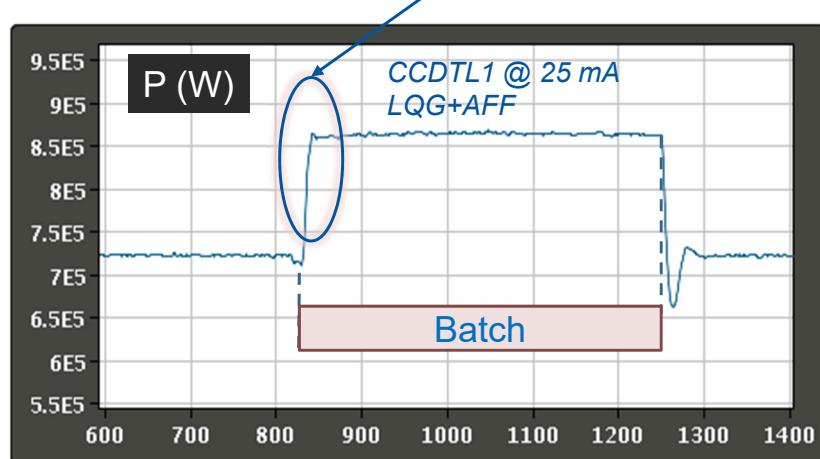


Power required

Large power overshoot



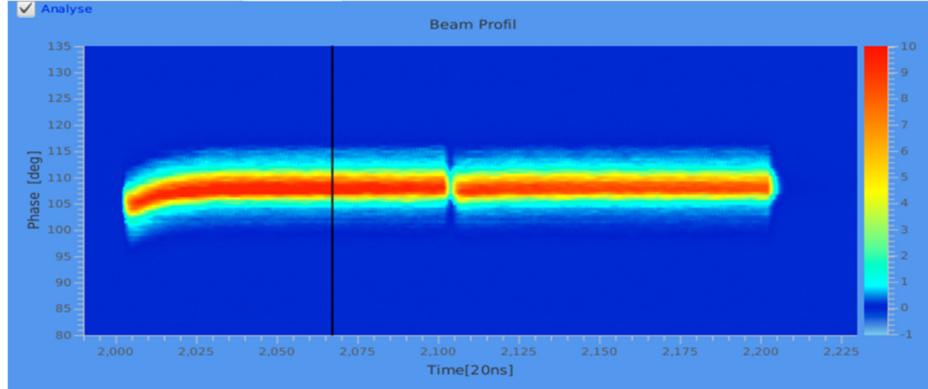
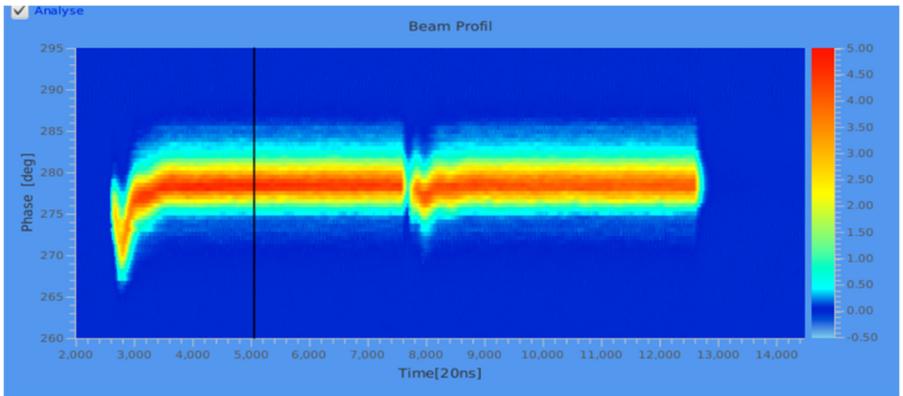
NO power overshoot



- The LQG regulation cannot anticipate the arrival of the head of the batch. It is caught by surprise and reacts violently causing a **peak of demanded power**
- With the AFF active the system **anticipates** and there is **no power surge** at the batch head
- The undershoot in power is present in both figures as the AFF is switched OFF at the end of the batch.



Bunch phase along batch(es)



Beam Phase Monitor measurement: Density plots showing the longitudinal position, within the reference 352.2 MHz buckets

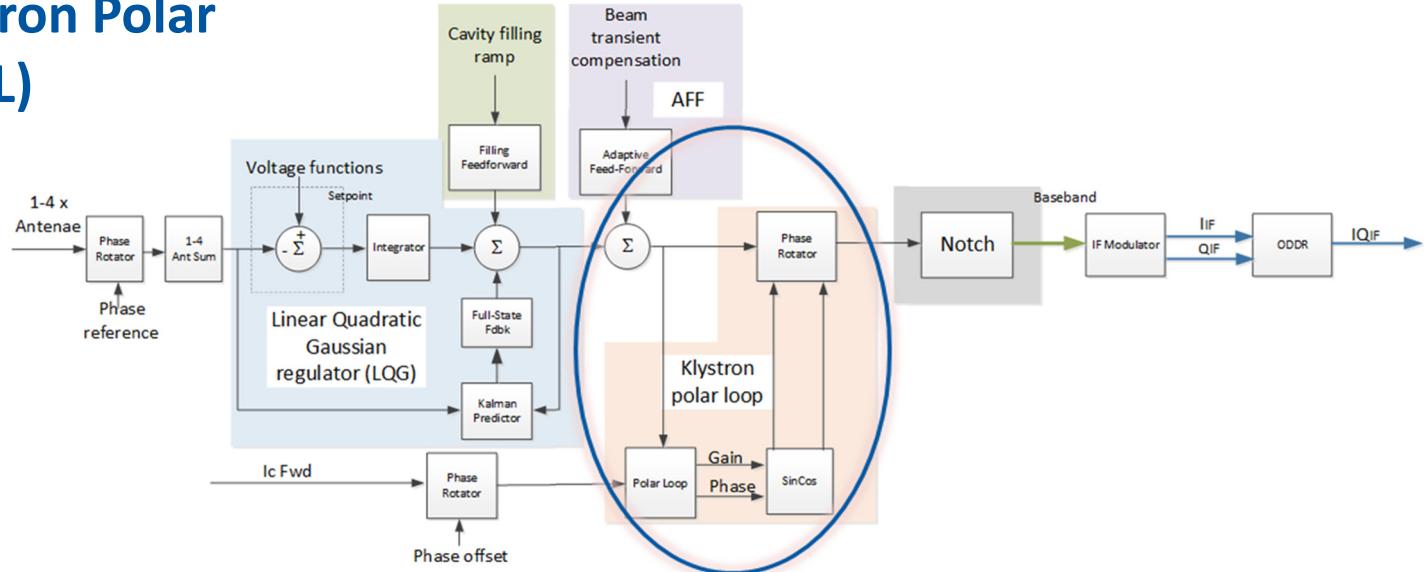
- Two batches
- Left: LQG only
- Right: LQG + AFF

Courtesy of Piotr Skowronski, CERN BE-OP





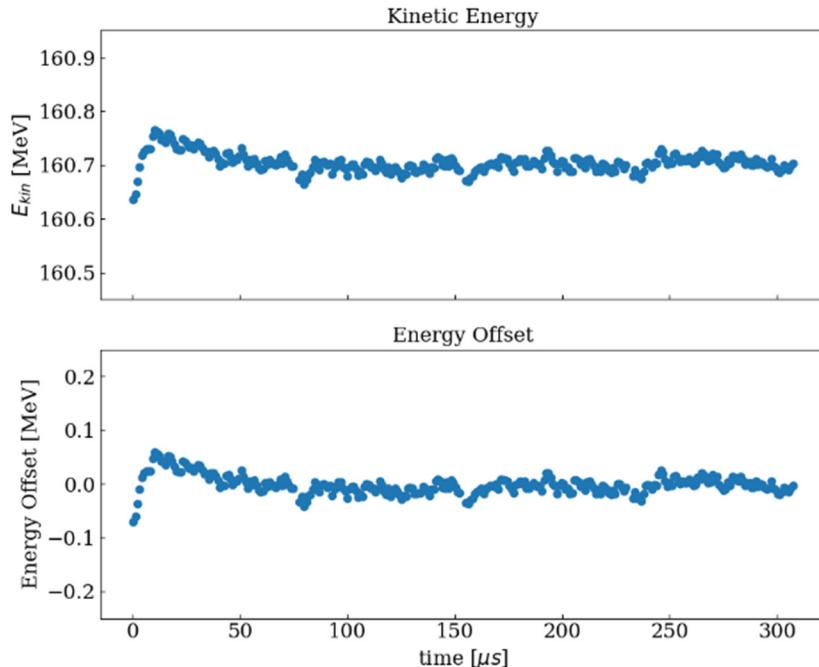
The Klystron Polar Loop (KPL)



- Another noise is the modulation of amplifier gain and phase, consequence of the ripples on the High Volt-age supply
- Multiplicative noise, therefore best compensated by acting on the LLRF gain and phase shift (polar)
 - We compare the forward current at cavity input $I_{c,fwd}$ (measured with a coupler in the waveguide before the cavity main coupler) to the sum of LQG and AFF outputs
 - We apply gain and rotation to keep the overall gain and phase shift constant, including klystron and circulator.
 - The reaction time is chosen to be much slower than the LQG, but sufficient to cover slow HV drifts and the rectifiers ripples measured around 10 kHz.



Conclusions



- The field regulation relies on the **LQG**, the **AFF** and the **KPL**
- The plots show **beam energy measurements along the batch** with four 75 μ s long batches spaced by 2 μ s, 21 mA DC
- The energy measured at Linac4 output deviates by **± 50 keV pk-pk** around the 160.7 MeV design value
- That is within specs, given the **design rms energy spread** of the individual 352.2 MHz bunch, that is **250 keV**
- It will still be acceptable extrapolating to the future 40 mA DC current.

Thank you for your attention
Questions? Comments?



References

- [1] Linac4 Technical Design Report, CERN-AB-2006-084 ABP/RF <https://cds.cern.ch/record/1004186/files/ab-2006-084.pdf>
- [2] G. Bellodi, M. Eshraqi, M. Garcia Tudela, L. Hein, J.B. Lallement, S. Lanzone, A.M. Lombardi, P. Posocco, E. Sargsyan, Alignment and Field Error Tolerance in Linac4 , CERN-ATS-Note-2011-021 <https://cds.cern.ch/record/1342092/files/CERN-ATS-Note-2011-021.pdf>
- [3] J.Tuckmantel, Cavity-Beam-Transmitter Interaction Formula Collection with Derivation, CERN-ATS-Note-2011-002 TECH , Jan 2011, <https://cds.cern.ch/record/1323893/files/CERN-ATS-Note-2011-002%20TECH.pdf>
- [4] C. Ziomek and P. Corredoura, "Digital I/Q Demodulator", in Proc. PAC'95, Dallas, TX, USA, May 1995, paper RPQ02
- [5] P. Baudrenghien, J. Galindo, G. Hagmann, J. Noirjean, D. Stellfeld, D. Valuch, COMMISSIONING OF THE LINAC4 LOW LEVEL RF AND FUTURE PLANS, Linac 2014 conference <http://cds.cern.ch/record/2062609/files/thpp027.pdf>
- [6] G.F. Franklin et al., Digital Control of Dynamic Systems, Ellis-Kagle Press, 3rd edition, section 8.3, pp302-310
https://www.researchgate.net/publication/31849881_Digital_Control_of_Dynamic_Systems-Third_Edition
- [7] B. Bielawski, P. Baudrenghien, R. Borner, RECENT DEVELOPMENTS IN LLRF FOR CERN'S LINAC4, LLRF 2019 workshop, <https://arxiv.org/pdf/1909.12541.pdf>
- [8] M. Laverty and K. Fong, An Iterative Learning Feedforward Controller for the TRIUMF e-linac, Proc. of LINAC2016, East Lansing, MI, USA
<https://accelconf.web.cern.ch/linac2016/papers/tuplr009.pdf>

