DESIGN OF CAVITY BPM PICKUP FOR EuPRAXIA@SPARC_LAB

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Introduction

- We have designed a Cavity Beam Position Monitor for the EuPRAXIA@SPARC_LAB project.
- A particular set of parameters for the prototype were determined in order to fulfil the EuPRAXIA@SPARC_LAB requirements.
- Based on this set, simulations were performed with ANSYS HFSS for frequency domain calculations, using the wire technique, and CST for simulating the passing particle beam, to verify the goodness of the design layout and optimise it.
- Sensitivity to the possible fabrication errors and performance degradation of the prototype have been studied.
- Theoretical resolution of the device was approximated with the upper limit for additional contribution in noise due to the mechanical tolerances.
- Mechanical implementation, based on the new clamping method, developed at LNF INFN, is proposed.
- Foreseen future tests.

EuPRAXIA@SPARC_LAB project and it's requirements



The layout of the EuPRAXIA@SPARC_LAB infrastructure.

The control of the charge and the trajectory at a few pC and few μ m is mandatory in this machine, especially in the plasma interaction region. However, other kind of BPMs, such as stripline BPM, can be used only at the beginning of the accelerator, where the beam pipe is 40 mm. But starting from X band structures, the pipe size decreases. Also, one of the most important parameter was considered to be the length of the device.



EuPRAXIA@SPARC_LAB project beam specifications.

Parameter	Units	Full rf	LWFA	PWFA
Electron energy	${ m GeV}$	1	1	1
Repetition rate	Hz	10	10	10
RMS Energy Spread	%	0.05	2.3	1.1
Peak Current	kA	1.79	2.26	2.0
Bunch charge	m pC	200	30	200(D)-30(W)
RMS Bunch Length	$\mu \mathrm{m}\mathrm{(fs)}$	16.7 (55.6)	2.14(7.1)	3.82(12.7)
RMS normalized Emittance	$\operatorname{mm} \operatorname{mrad}$	0.05	0.47	1.1
Slice Length	$\mu { m m}$	1.66	0.5	1.2
Slice Charge	\mathbf{pC}	6.67	18.7	8
Slice Energy Spread	%	0.02	0.015	0.034
Slice normalized Emittance (x/y)	$\operatorname{mm}\operatorname{mrad}$	0.35/0.24	0.45/0.465	0.57/0.615
Undulator Period	mm	15	15	15
Undulator Strength $K(a_w)$		0.978~(0.7)	1.13(0.8)	1.13(0.8)
Undulator Length	m	30	30	30
$ ho ~(1\mathrm{D}/3\mathrm{D})$	$\times 10^{-3}$	1.55/1.38	2/1.68	2.5/1.8
Radiation Wavelength	nm (keV)	2.87(0.43)	2.8(0.44)	2.98(0.42)
Photon Energy	$\mu { m J}$	177	40	6.5
Photon per pulse	$ imes 10^{10}$	255	43	10
Photon Bandwidth	%	0.46	0.4	0.9
Photon RMS Transverse Size	$\mu{ m m}$	200	145	10
Photon Brilliance per shot	(s mm2 mrad2 bw(0.1%)) $^{-1}$	1.4×10^{27}	$1.7{\times}10^{27}$	0.8×10^{27}

EuPRAXIA@SPARC_LAB's beam parameters for plasma and conventional RF linac driven FEL.

Dual-Resonator cBPM full object sketches



Cavity BPM pickup schematic view (shown: vacuum).

Parameters	Values
Working frequency range	C band
Loaded quality factor Q_L	≈ 500
Sensitivity	$\approx 5 \text{ V/nC/mm}$
Required resolution	$< 1 \mu { m m}$



Cavity BPM pickup half-cut schematic view (shown: copper shell).

Sensitivty of the device should be of the order of 5 V/nC/mm in order to provide <1 μ m resolution.

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Working principle and theory of Cavity Beam Position Monitors.



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Geometrical parameters for the monitor.





Transit time factor for the monopole and dipole modes of a pillbox cavity with different lengths.

Dependence of the shunt impedance on the cavity length.



The width of waveguide (a) has to be chosen such, that its cut-off frequency is located between TM_010 and TM_110 cavity modes.

The monopole signal is exponentially decaying along the waveguide, therefore, it is better to have minimal height (b) (compromise has to be found between height and coaxial output transmission quality).

The length (c) has to be chosen in order to eliminate reflections.

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Waveguide to coaxial transition.



Different types of coupling were simulated for waveguide to coaxial transition substructure.





Exposition of how sufficient transition is achieved between the waveguide and coaxial antenna by changing the x and z parameters.

Waveguide to coaxial transition for dual-resonator cBPM.





dB S(1,1) parameter for 5.1 GHz waveguide to coaxial transition for the direct coupling with whip antenna.



Mag S(1,2) parameter for 5.1 GHz waveguide to coaxial transition for the direct coupling case.

dB S(1,1) parameter for 5.1 GHz waveguide to coaxial transition for the inductive coupling with whip antenna.

dB S(1,1) parameter for 5.1 GHz waveguide to coaxial transition for the capacitive coupling.

Dimmension	Value/mm		
	Direct Inductive		Capacitive
	$\operatorname{coupling}$	$\operatorname{coupling}$	$\operatorname{coupling}$
Whip Antenna Raduis	0.635	0.635	0.635
Bead Radius	—	—	1.8
Bead Height h $_{bead}$	—	—	1.2
Curvature Radius R $_{curve}$	—	0.5	0.2
Spacing s	0	0.151	1.969
Distance from the short-end z	19.5	58.8	21.74
Distance from the wall x	25	18.5	12.2
Waveguide height h $_{wg}$	3	8	6
Waveguide height w $_{wg}$	39	37	37
Waveguide height l $_{wg}$	57	90	57

E Field [V/n] mag S(1.2) and mag S(1.3) 0 0 0 7 7 0 8 0 8 14.1683 7.6674 9.1499 2.2455 1.2152 0.6576 0.3559 0.1926 0.1042 0 3 5 6 4 7 Frequency [GHz] E Field [V/m N6.5412 87.5375 81.4095 54.9231 29.7227 16.0550 8.7047 4.7187 2.5493 1.3796 8.746 0.4040 0.2187 0 3 5 6 4 7 Frequency [GHz]

Simulting position cavity with wire for dual-resonator cBPM.

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Simulting position cavity with wire for dual-resonator cBPM.

Optimizing position cavity length for dual-resonator cBPM.





Output voltage V_{out} dependence on the position cavity length variation.

After the prototype conceptual design was obtained, separate parts still need to be tuned to match the required properties and deliver optimal performance. However, the cavity beam position monitor can be considered as a system, where one component/dimension variation causes changes to other parameters, one can still divide the whole system into separate parts and start the design process to unite them in one particular layout then.

To determine the optimum position cavity length, which will provide sufficient output voltage for the required spatial resolution, corresponding simulation and long enough decay time τ , the output voltage dependence on the position cavity length with quality factor was set up.

Optimizing WG and **X**, **Z** dimensions for Dual-Resonator cBPM.



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Position cavity's dimensional values and tolerances.

The 5.1 GHz prototype position cavity's optimised design.

Dimmension	Value[mm]
Position Cavity Radius R	34.07
Position Cavity Length L	5
Waveguide Length WG length	21.57
Waveguide Width WG width	22.6
Waveguide Height WG height	3
Coaxial Position x	6.3
Coaxial Position z	7

Dimension values for the 5.1GHz prototype position cavity.

1.2323E+13 3.6888E+12					
1.0569E+12 3.0953E+11 9.0650E+10			Dimmension	Q_0 Variation	Frequency shift
2.6548E+10 7.7750E+09 2.2770E+09 6.6688E+08	A state	<u>104</u>		[%/mm]	[MHz/mm]
1.9530E+08 5.7198E+07 1.6751E+07	- And		Position Cavity Radius R	1	137.77
4. 9057E+06 1. 4367E+06			Position Cavity Length L	12	-28.06
4. 6353E+08 4. 3263E+08			Waveguide Length WG _{length}	-5	-4.2
4.0173E+08 3.7082E+08 3.3992E+08 3.0902E+08			Waveguide Width WG width	-0.87	-0.12
2.7812E+08 2.4722E+08 2.1631E+08	• • • • • • • • • • • • • • • • • • •		Coaxial Position x	13	-0.48
1.5451E+08 1.2361E+08 9.2706E+07			Coaxial Position z	9	0.45
3. 095(47) 3. 075(47) 3. 075(47)		Dimension tolerances to the 1mm change for the 5.1GHz prototype position cavity.			

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Poynting [W/m^2]

1.4367E+14 4.2076E+13

Reference cavity's dimensional values and tolerances.

Dimmension	Value	Deviation,	Q variation
		[MHz/1 mm]	$[\%/\mathrm{mm}]$
Reference Cavity Radius R	22.71	-221	5.6
Reference Cavity Length L	9	25.43	20
Trench Radius r	21.71	-315.38	-26
Reference Cacity Effective Length G	4	-175	-22
Distance between cavity wall and coaxial coupler D	2.6	3.5	-30

Dimensions and tolerances to 1 mm change for the 5.1GHz prototype reference

cavity.

Schematic view of reference cavity with geometrical indications.

It consists of a particular pillbox where the antenna is inserted in the bulge. In general, the insertion depth of the antenna in the bulge determines the desired output signal level. In particular, the closer the antenna end is to the cavity wall, the higher the output signal amplitude becomes.

Output voltage signal coming out from the reference cavity with a 1 mm beam offset. Wakefield simulation.

Dual-Resonator cBPM prototype wakefield simulations.

Output voltage signal coming out of one of the coaxial ports for position cavity.

FFT of the signal coming out from the coaxial port.

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Sensitivity and mechanical tolerances of the monitor.

Sensitivity and theoretical res. for dual-resonator cBPM.

Position cavity ports cross-coupling and isolation between the position and reference cavities.

Thermal noise level.

Additional 10 μ V contribution in total V_{out} due to fabrication errors. Additional 20 μ V contribution in total V_{out} due to fabrication errors. Additional 30 μ V contribution in total V_{out} due to fabrication errors. Output voltage due to the beam offset in mm.

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Mechanical implementation.

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Assembled prototype.

The device is assembled from three parts, using the clamping method.

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Future tests.

Once the first prototypes will be fabricated, bench-top measurements and beam tests will be performed. Such a new test-bench for cBPMs at SPARC-LAB at INFN-LNF was designed and will be used in the prototype tests.

The test bench aims to perform measurements on the manufactured cBPMs. The main reason for these is to investigate further the prototype presented in this poster and its properties, dealing with the new challenges related to beam diagnostics for the EuPRAXIA@SPARC_LAB.

Conclusions

- Cavity BPM design process and strategy for achieving the required specifications are described.
- **Dual-resonator cavity BPM** prototype is proposed.
- Sensitivity and theoretical resolution for developed cBPM are evaluated.
- Future tests, once the monitor is manufactured, are determined.

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