



Beam Physics Limitations for Damping of Instabilities in Circular Accelerators

Valeri Lebedev

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Objectives

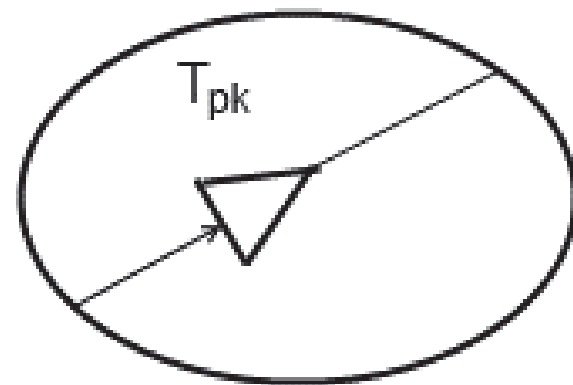
- To discuss delusions which disappeared with more work
 - ◆ Some delusions were discarded during this talk preparation
- To share accumulated experience

Talk Outline

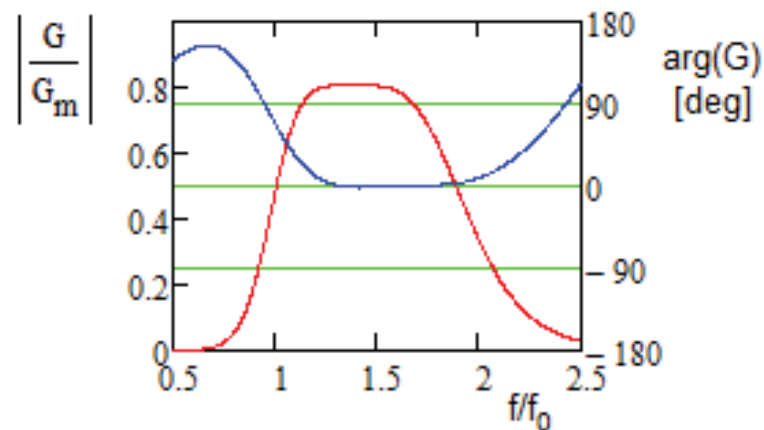
- Causality and correction of damper transfer function
- Emittance growth suppression by a damper
- Limitations on the FB system gain
- Analog preprocessing and postprocessing in digital FB systems
- Effects of x-y coupling on damping

Causality

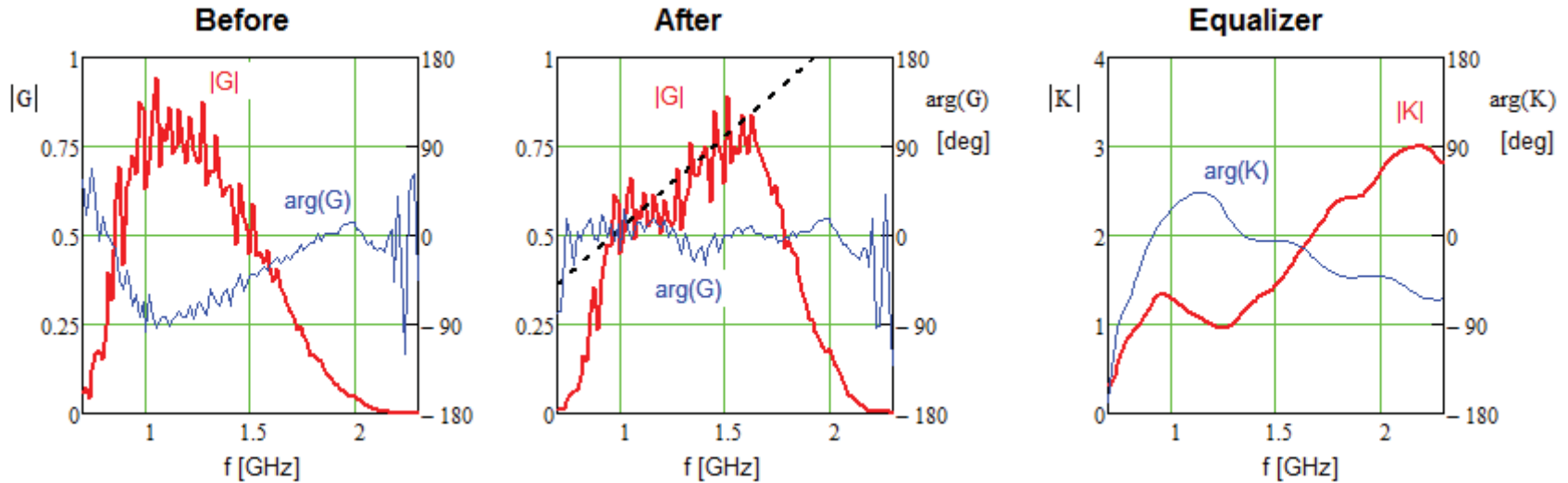
- Causality results in that for typical amplifier the amplitude and phase responses are related (Kramers - Kronig relations)
 - ◆ However, there is no causality limitation in a damper
 - ◆ Shorter cable can make signal coming ahead of bunch (particle)
- ⇒ Amplitude and phase responses can be controlled independently but it requires additional time
- Main limitations for digital filtering are the same but digital filtering adds additional flexibility to a system



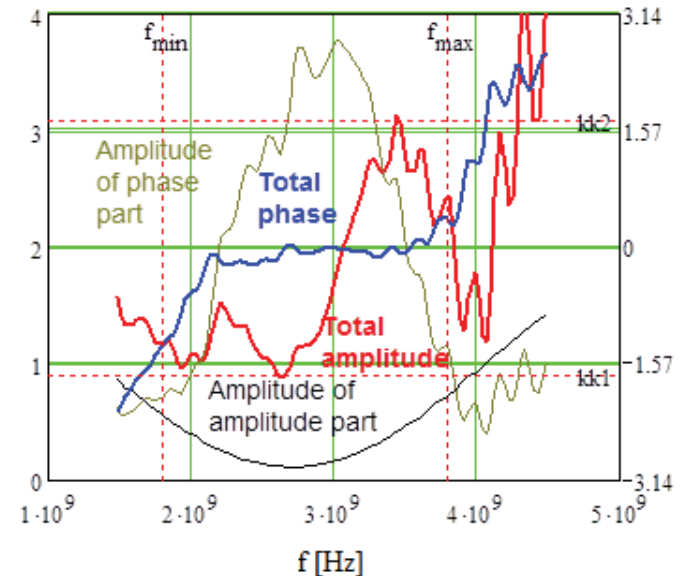
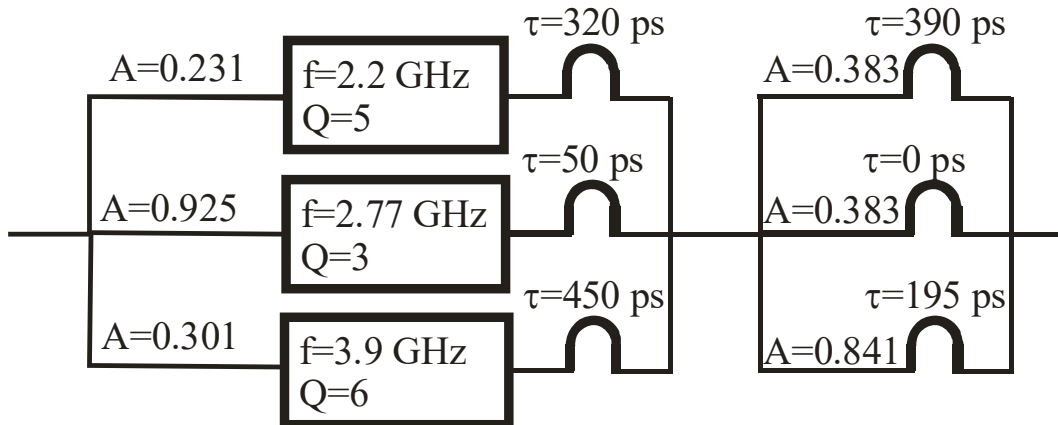
Phase in the band center is corrected by negative delay



Equalizers for Stochastic Cooling



Correction of transfer function for Recycler



Equalizer for Accumulator Stacktail

Analog Circuit to Create $1/\sqrt{\omega}$ Gain Dependence

- Often $\perp$ multi-bunch instability is driven by resistive wall impedance

$$\Rightarrow \lambda \propto 1/\sqrt{\omega}$$

- That determined the choice of this example

$$\begin{aligned} \tau_1 &:= 1 & \tau_2 &:= 0.02 \\ \tau_3 &:= 0.02 & \tau_4 &:= 0.004 \\ \tau_5 &:= 4 \cdot 10^{-4} & \tau_6 &:= 8 \cdot 10^{-6} \end{aligned}$$

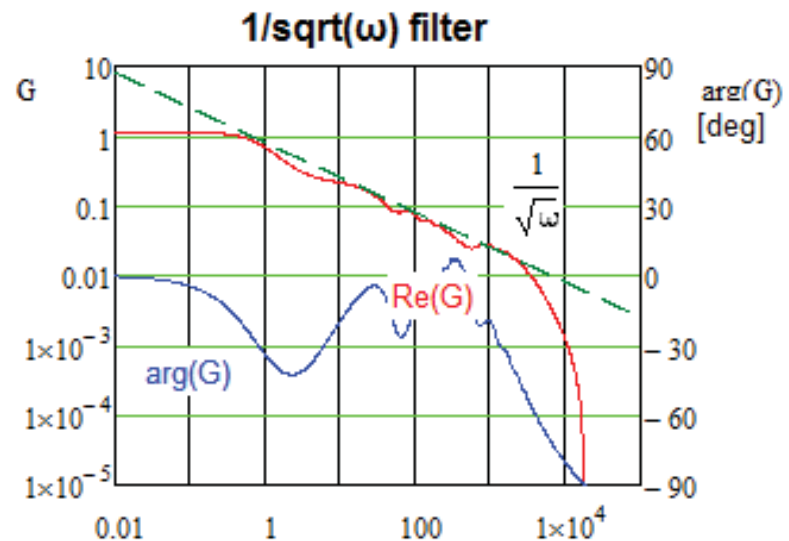
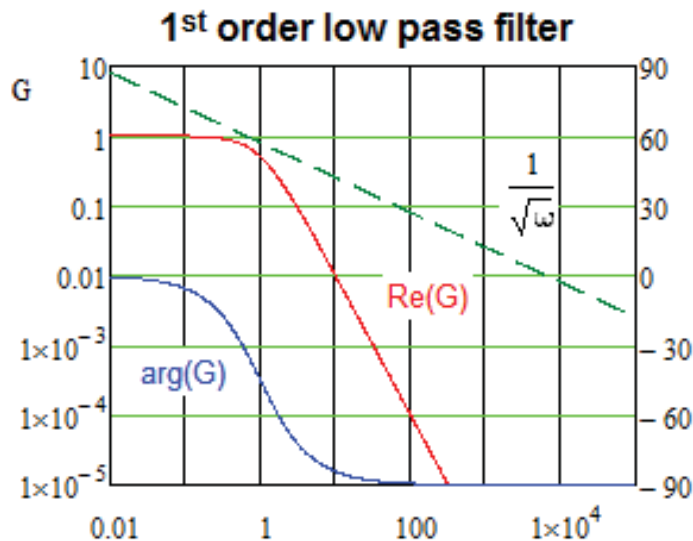
$$G(\omega) := \frac{1}{1 + i\omega\tau_1}$$

$$G_{c1}(\omega) := \frac{\exp(-i\omega\tau_1)}{(1 + i\omega\tau_1) \cdot (1 + i\omega\tau_2)}$$

$$G_{c2}(\omega) := \frac{0.11 \cdot \exp(0.01i\omega\tau_1)}{(1 + i\omega\tau_3) \cdot (1 + i\omega\tau_4)}$$

$$G_{c3}(\omega) := \frac{0.03 \cdot \exp(0.00i\omega\tau_1)}{(1 + i\omega\tau_5) \cdot (1 + i\omega\tau_6)}$$

$$G_c(\omega) := G_{c1}(\omega) + G_{c2}(\omega) + G_{c3}(\omega)$$



Right: an example of analog filter with $1/\sqrt{\omega}$ gain

- Method can be used for correction the power amplifier phase response
 - At sufficiently low frequencies it can be done with digital filter

Suppression of Emittance Growth by FB

- Without damping, $\perp$ kicks ($\Delta\theta$) result in growth of betatron amplitude

$$\frac{d}{dt} \overline{\Delta x^2} = \frac{f_0}{2} \beta^2 \overline{\Delta \theta^2}$$

- For noise with spectral density $S_\theta(\omega)$ we have

$$\frac{d}{dt} \overline{\Delta x^2} = \frac{\beta^2 \omega_0^2}{4\pi} \sum_{n=-\infty}^{\infty} S_\theta((\nu - n)\omega_0), \quad \omega_0 = 2\pi f_0$$

where the normalization is: $\overline{\Delta \theta^2} = \int_{-\infty}^{\infty} S(\omega) d\omega$

- Beam decoherence results in emittance growth

$$\left(\frac{d\varepsilon}{dt} \right)_0 = \frac{\omega_0^2}{4\pi} \sum_k \beta_k \sum_{n=-\infty}^{\infty} S_{\theta_k}((\nu - n)\omega_0) \xrightarrow[\text{single source}]{\text{white noise}} \frac{f_0 \beta}{2} \overline{\Delta \theta^2}$$

Only resonant frequencies contribute to the emittance growth

- If decoherence is slower than damping, $d\varepsilon/dt$ is suppressed^[1]:

$$\frac{d\varepsilon}{dt} = \frac{16\pi^2 \overline{\Delta \nu^2}}{g^2} \left(\frac{d\varepsilon}{dt} \right)_0, \quad g \gg \sqrt{\Delta \nu^2}$$

where the damping rate in amplitude is: $\lambda = f_0 g / 2$

and $\sqrt{\Delta \nu^2}$ is the rms tune spread

[1] Particle Accelerators, 1994, Vol. 44, pp. 147-164 and pp. 165-199

Emittance Growth due to FB System Noise

- Noise in FB system results in additional excitation of betatron oscillations

- ◆ Referencing all FB system noises to the BPM one obtains:

$$\frac{d\varepsilon}{dt} \approx \frac{16\pi^2 \overline{\Delta v^2}}{g^2 + 16\pi^2 \overline{\Delta v^2}} \left[\left(\frac{d\varepsilon}{dt} \right)_0 + \frac{f_0 g^2}{2\beta_{pk}} \sigma_{pk}^2 \right], \quad \sigma_{pk} = \sqrt{X_{BPM}^2}$$

- Suppression of BPM noise by gain reduction with frequency

- ◆ $g \gg \sqrt{\Delta v^2}$ then effect of FB noise does not depend on gain
- ◆ If at high freq. $g \leq \sqrt{\Delta v^2}$ and external noise is negligible then the BPM noise is suppressed. Actual noise reduction depends on parameters.

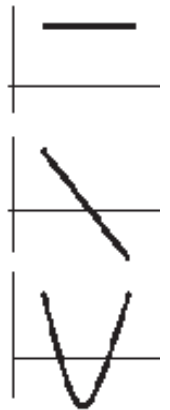
$$\frac{d\varepsilon}{dt} \approx \sum_{n=-n_b/2}^{n_b} \frac{16\pi^2 \overline{\Delta v^2}}{g_n^2 + 16\pi^2 \overline{\Delta v^2}} \left[\left(\frac{d\varepsilon}{dt} \right)_n + \frac{f_0 g_n^2}{2\beta_p} \frac{\sigma_{bpm}^2}{n_b} \right]$$

- For head-on collisions in the collider: $\sqrt{\Delta v^2} \approx 0.2\xi$

- External noise is at low frequencies

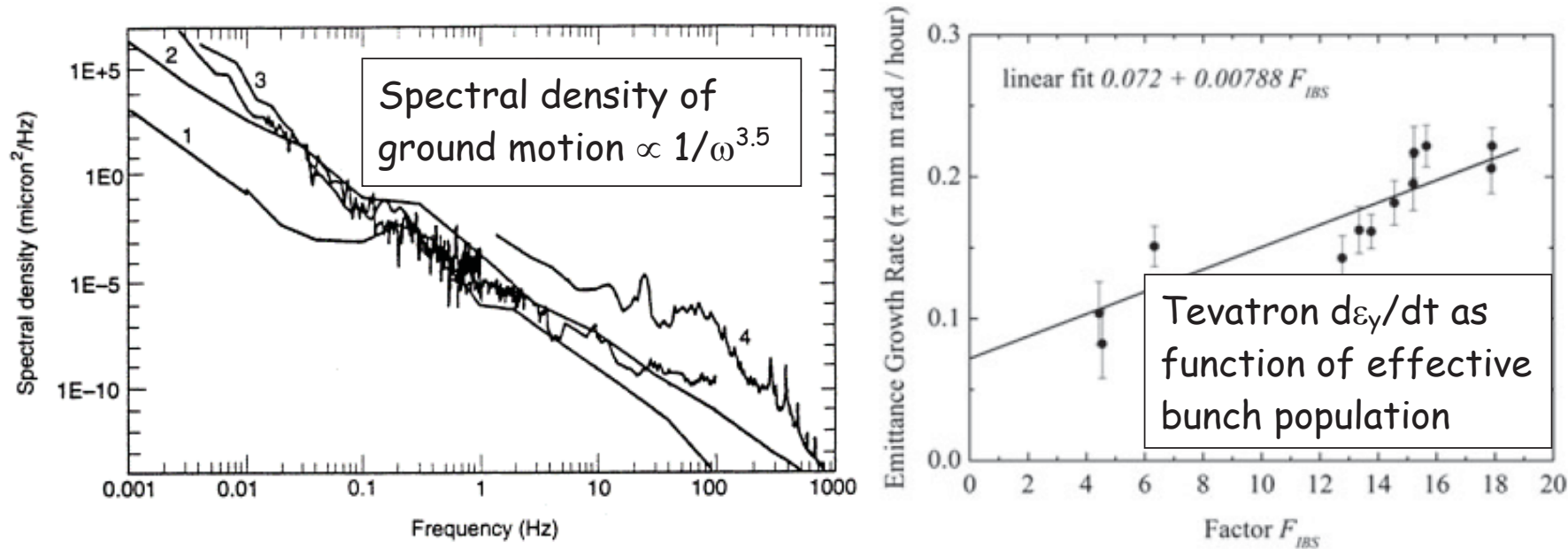
=> bunch is kicked as one whole

- ◆ For non-zero chromaticity synchrotron motion changes $\perp$ bunch shape
=> $g > \nu_s$ to prevent suppression of emittance growth



Sources of Emittance Growth

- Fluctuations of bending field
 - ◆ Fluctuations of current
 - ◆ Oscillations of liners inside SC dipoles (frozen B field), $\Delta B/B \ll 10^{-9}$
- Quad displacement due to ground motion,
- In Tevatron at collisions about 20% of emittance growth is related to field fluctuations
 - ◆ Inability to operate at low betatron tune
 - ◆ Emittance growth due to scattering on residual gas are excluded by other measurements



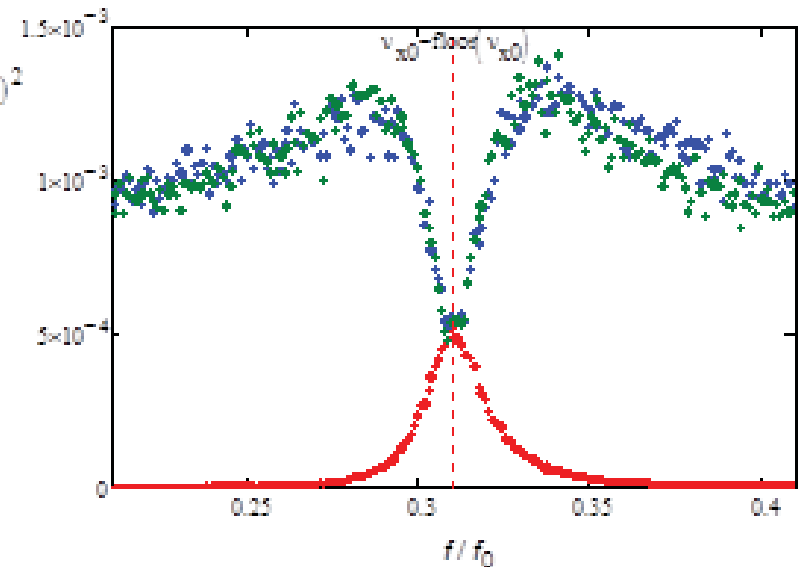
Challenges of Large Hadron Colliders

- With energy increase the revolution frequency comes to audio-frequencies where noise spectral density is high
- LHC “hump” (summer of 2010) resulted in unacceptably large emittance growth + large jumps of emittances
 - ◆ These harmful effects were suppressed by gain increase in transverse dampers ($g \gg \xi$)
 - ◆ It also required a reduction of damper noise
 - Achieved accuracy of BPMs of about $\sim 0.2 - 0.5 \mu\text{m}$ still produced measurable emittance growth
 - ◆ Noisy power supplies were found in about half year
- Problems will grow fast with an increase of machine energy due to coming to even smaller frequencies
- Large size excludes using analog system (digital notch filter)

$$\delta\theta_n = \frac{g_1}{\sqrt{\beta_{p1}\beta_{kick}}} \sum_{k=0}^{K-1} A_k (x_{n-k} + \delta x_{n-k}), \quad \xrightarrow[\text{condition}]{\text{notch filter}} \sum_{k=0}^{K-1} A_k = 0$$

Suppression of BPM Noise by Digital Filters

- Can optimally built digital filter reduce sensitivity to BPM errors?
 - ◆ **The answer is no!!!**
 - Simple explanation is that each error is applied K times. Errors are added coherently the same as damping terms.
 - This final answer is correct if the damper is in the linear regime. I.e. the gain is sufficiently small so that FB system is far from instability
 - Formal prove is in: W. Hofle, $(13)^2$
V. Lebedev et al. IPAC'11
- Measurements of spectrum of BPM signal enables computation of betatron frequency, and damper gain and phasing
- More points are used in filter more sensitive is the damper to a betatron frequency error and smaller maximum gain is achievable $g_{max} \propto 1/K$



*Spectral density of noise for
two-BPM LHC system
Red - actual beam motion*

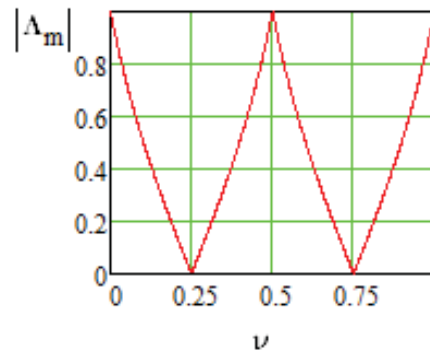
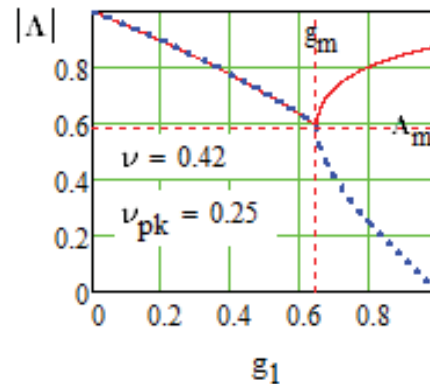
Limitations on the FB System Gain

- Large hadron colliders require large FB system gain to suppress $d\varepsilon/dt$

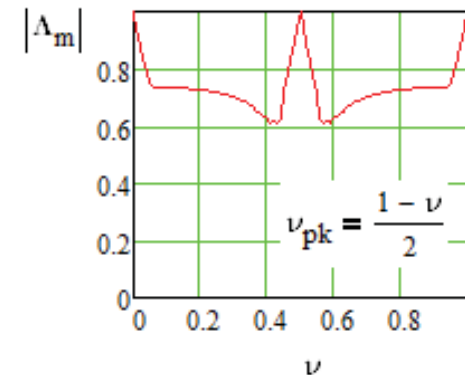
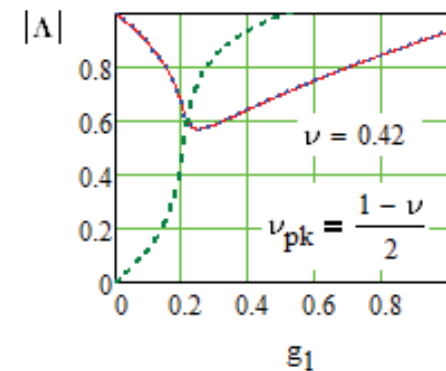
$$\mathbf{x}_{n+1} = \mathbf{M}_{kp} \left(\mathbf{M}_{pk} + \begin{bmatrix} 0 & 0 \\ 0 & g \end{bmatrix} \sum_{k=0}^{K-1} A_k \mathbf{x}_{n-k} \right) \Rightarrow \left| \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \lambda - \mathbf{M} - \mathbf{M}_{kp} \begin{bmatrix} 0 & 0 \\ 0 & g \end{bmatrix} \sum_{k=0}^{K-1} A_k \lambda^{-k} \right| = 0$$

- Usage of larger number of turns for correction computation reduces achievable damping rate;

$$\lambda_{max} \propto 1/K.$$



One-turn system (no notch filter): optimal $\nu_{pk}=0.25$, 2 eigen-values



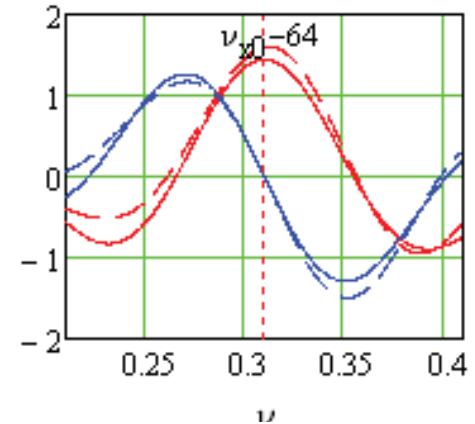
Two-turn system (notch filter): opt. $\nu_{pk}=(1-\nu)/2$, 4 eigen-values ($\Lambda_0=0$, 3 others are shown)

LHC Transverse Dampers (as in 2011)

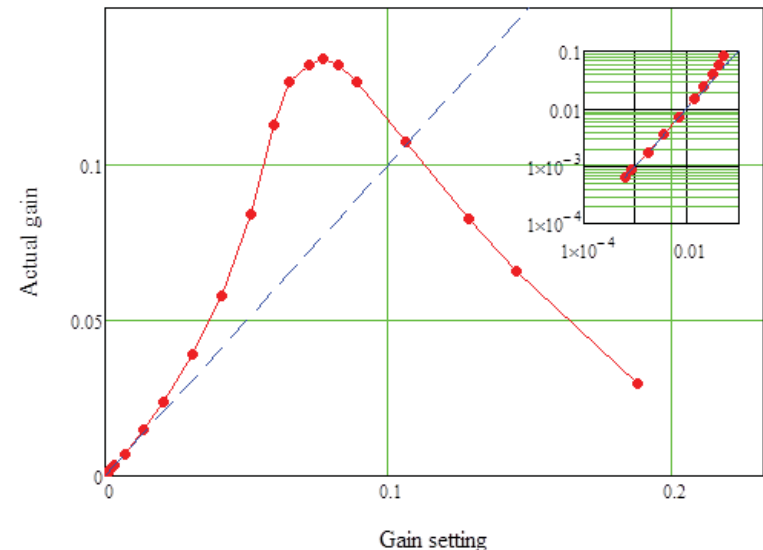
- In 2011 the LHC transverse damper was based on 7th order filter

$$A_k = \begin{bmatrix} -\frac{2}{3\pi} \sin \psi & 0 & -\frac{2}{\pi} \sin \psi & \cos \psi & \frac{2}{\pi} \sin \psi & 0 & \frac{2}{3\pi} \sin \psi \end{bmatrix}^T$$

- That significantly reduced the maximum achievable gain of the system and introduced excessive sensitivity to the machine tune
- As far as we understand now such choice did not deliver any advantages
 - ◆ In particular, same sensitivity to the BPM noise
- “Reasonable” filter should use 3 turns to accommodate notch filter and arbitrary phase advance between pickup and kicker



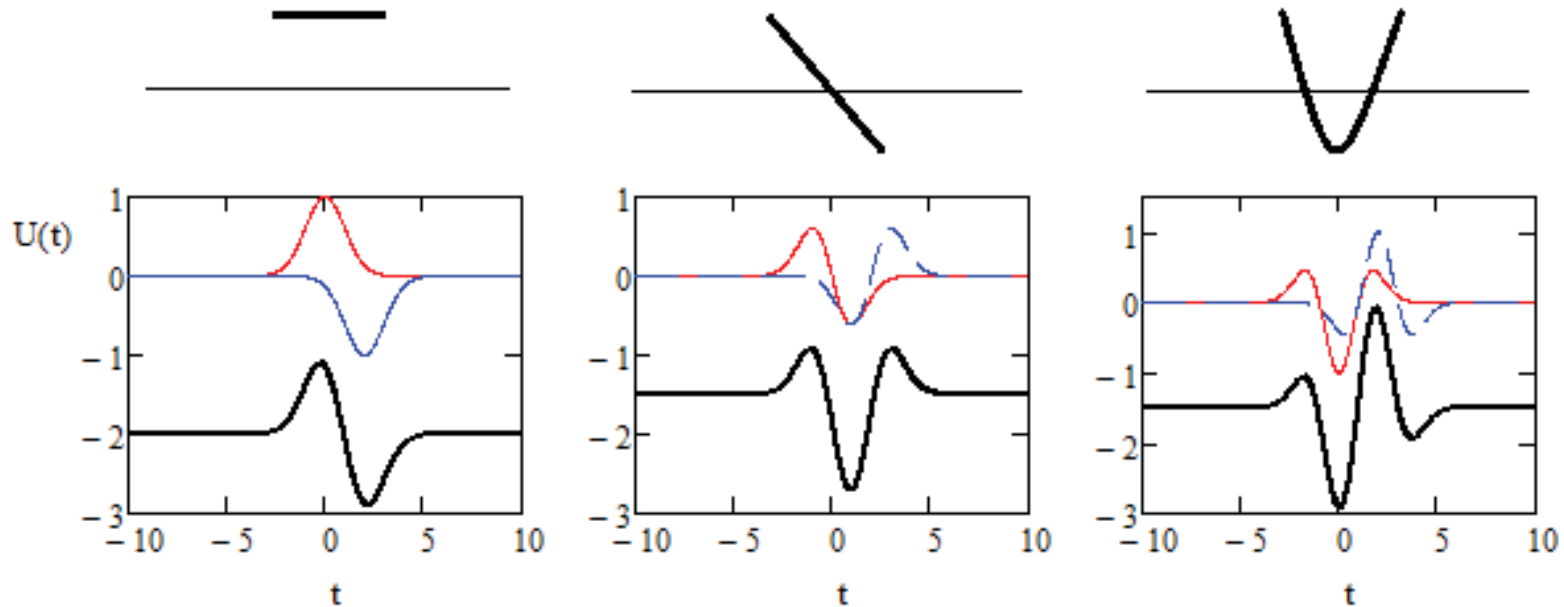
Real (red) and imaginary (blue) parts of the gain on the machine tune



What does a BPM Measures?

■ Strip-line BPM out voltage

$$U(t) = Z_{cpl} (I(t) - I(t - 2L / c)), \quad v = c$$



■ BPM signals are similar for

- ◆ intra-bunch HOM &

development of bunch betatron oscillations after uniform bunch kick

■ Result of BPM measurements depends on signal treatment before digitization (analog preprocessing)

- ◆ If $\sigma_b \ll$ bunch-bunch distance, an analog integration yields center of gravity

Analogue Preprocessing & Postprocessing in Digital FB

■ The following analogue preprocessing methods are usually used:

- Integration. It may deliver the center of gravity ($\sigma_b \ll L_{bb}$)
- Mixing with RF + low pass filter ($\sigma_b \ll L_{bb}$)
- Excitation of oscillator with subsequent digitization's ($\sigma_b \ll L_{bb}$)
- More than one-point digitization ...

◆ Typically all methods may be sensitive to HOMs

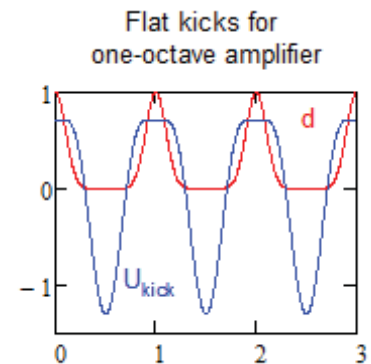
■ Similar the kick value across bunch may depend on time

- ◆ That makes non-uniform kicks
- ◆ May result in additional emittance growth and excitation of HOMs

■ Thus, analog preprocessing and post processing affect on the HOM damping/excitation

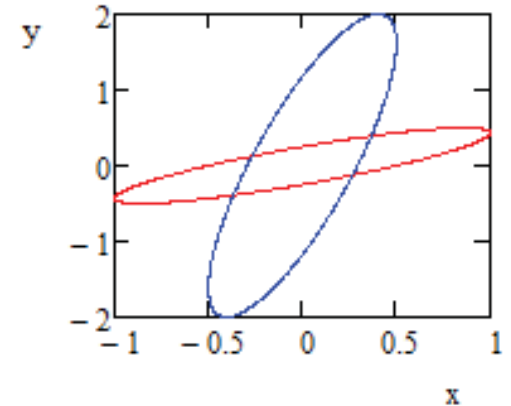
In 1st order PT: $\lambda_n \propto -\int X_n(s)U_n(s)ds$

- ◆ It limits the gain for zero (dipole) mode due to excitation of HOMs
- ◆ Correctly chosen analog preprocessing and post processing result in a reduction of HOMs excitation and, possibly, damping for some of them
 - It depends on details of each machine
 - It is an area requiring further studies



Effects of X-Y Coupling

- In the course of Tevatron Run II we observed that switching on a one-plane damper could introduce instability in another plane
- The reason of such behavior was strong x-y coupling which could not be completely compensating because of uncontrolled skew-quad components in SC dipoles
- Running dampers for both planes made beam stable



The analysis of the problem can be done similar to a single dimensional case where 2D matrices are replaced by 4-D

$$\mathbf{x}_{n+1} = \mathbf{M}_{kp} \left(\mathbf{M}_{pk} + \mathbf{G}_{x,y} \sum_{k=0}^{K-1} A_k \mathbf{x}_{n-k} \right) \Rightarrow \left| \lambda \mathbf{I} - \mathbf{M} - \mathbf{M}_{kp} \mathbf{G}_{x,y} \sum_{k=0}^{K-1} A_k \lambda^{-k} \right| = 0$$

$$\mathbf{I} = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}, \quad \mathbf{G}_x = \begin{bmatrix} 0 & 0 & 0 & 0 \\ g & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}, \quad \mathbf{G}_y = \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & g & 0 \end{bmatrix}$$

Perturbation Theory for Symplectic Motion

■ In many applications a solution with perturbation theory is sufficient

- ◆ Unperturbed motion $\mathbf{M}\mathbf{v}_j = \lambda_j \mathbf{v}_j$
- ◆ Perturbed motion $(\mathbf{M} + \Delta\mathbf{M})(\mathbf{v}_j + \Delta\mathbf{v}_j) = (\mathbf{v}_j + \Delta\mathbf{v}_j)(\lambda_j + \Delta\lambda_j)$
- ◆ Tune shifts:
$$\begin{bmatrix} \Delta\nu_1 \\ \Delta\nu_2 \end{bmatrix} = -\frac{1}{4\pi} \begin{bmatrix} \mathbf{v}_1^+ \mathbf{S} \Delta\mathbf{M} \mathbf{v}_1 \\ \mathbf{v}_2^+ \mathbf{S} \Delta\mathbf{M} \mathbf{v}_2 \end{bmatrix}$$

■ For damper we have

$$\left| \lambda \mathbf{I} - \mathbf{M} - \mathbf{M}_{kp} \mathbf{G}_{x,y} \sum_{k=0}^{K-1} A_k \lambda^{-k} \right| = 0 \quad \Rightarrow \quad \Delta\mathbf{M}_{1,2} = \mathbf{M}_{kp} \mathbf{G}_{x,y} \sum_{k=0}^{K-1} A_k \lambda_{1,2}^{-k}$$

For horizontal damper one obtains ($\mathbf{G}_x \rightarrow \mathbf{G}_y$ for vertical damper)

$$\begin{cases} \Delta\nu_1 = -\frac{1}{4\pi} \left(\sum_{k=0}^{K-1} A_k \lambda_1^{-k} \right) \mathbf{v}_1^+ \mathbf{S} \mathbf{M}_{kp} \mathbf{G}_x \mathbf{v}_1 \\ \Delta\nu_2 = -\frac{1}{4\pi} \left(\sum_{k=0}^{K-1} A_k \lambda_2^{-k} \right) \mathbf{v}_2^+ \mathbf{S} \mathbf{M}_{kp} \mathbf{G}_x \mathbf{v}_2 \end{cases}$$

Conclusions

- Beam physics considerations should be important part of damper design
 - ◆ Ignoring them may result in compromised performance and excessive cost

- Damping in the course of slip-stacking is another topic not discussed here but important for support of Fermilab neutrino program.
It is in a tomorrow morning talk:
 - ◆ “High Intensity Proton Stacking at Fermilab: 700 kW Running”
R. Ainsworth