SPLIT HIGHER ORDER MODES IN SUPERCONDUCTING CAVITIES*

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Abstract

Split resonances are a common appearance in superconducting cavities and are studied here on the specific examples of the TE11 and TM11- like dipole resonances in five-cell copper models of the Brookhaven ERL and BNL3 cavities. The BNL3 was designed to be suitable for the envisioned high-intensity projects at this laboratory. Achieving the required high current performance depends on avoiding beam-breakup instabilities by minimizing the Higher Order Modes (HOM) Q-values. This was attempted in the design phase and will be done with appropriate mode dampers in operation. The availability of a copper model provided the convenient opportunity to confirm the design and to study potentially nefarious high-*O* resonances. The appearance of split resonances impeded the HOM identification and the theoretical interpretation as ellipticity deformation is presented in this paper.

INTRODUCTION

Improvements of RHIC operation will depend to a large degree on the application of superconductivity (SC) Intrinsic to all SC cavities is the to RF cavities. appearance of higher order modes and the permanent need of their study. The splitting of resonance frequencies is a specific problem, complicating the HOM mode identification and preventing reliable *O*-measurements. The splitting is found already in normal conducting (NC) cavities and can be conveniently performed on the Niobium cavity of the Energy Recovery Linac (ERL) [1] or the copper cavity model for the BNL3 cavity [2] shown in Fig. 1. The BNL3 was designed to be suitable for the envisioned high-intensity projects at this laboratory, and experimentally confirming the design assumptions is mandated. Significant frequency deviations from the design values are expected due to beam tube end effects. The small resonance frequency splitting is here explained, using Mathieu functions, as fabrication-caused ellipticity deformation.

BNL3 DIPOLE RESONANCES

Cavity resonances are obtained from S_{21} transmission measurements with a network analyzer (NA) [3]. BNL3 cavity has simple type-N connectors used as probes in cell 1 and 2 together with stronger coupling fundamental power coupler (FPC) and pickup (PU) probes. Figure 2 shows the transmission in the first dipole pass band through the cavity from cell 1 to 5 and across from FPC to PU. Note the repetitive resonances in the signal across the cavity, presumably due to the tapered end tubes. Not explained is a stop pass filtering in the central range.

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Table 1: Dipole Resonances		
f_{even}	$f_{\rm odd}$	$\Delta \times 10^4$
821.26	821.78	6
827.25	828.09	10
843.5	844.29	9
866.13	867.10	11
885.91	886.47	6
906.7	~	
932.31	~	
954.17	954.48	3
987.90	988.05	1
1007.03	1007.16	1
1016.78	1016.94	1
1020.95	1021.03	<1



Figure 1: BNL3 Copper cavity.



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Split Dipole Resonances at 821 MHz

The lowest dipole resonance at 821 MHz represents the prototypical split HOM and was fully measured with different excitations as shown in Fig. 3. The highest Qvalue obtained is 21,300, essentially independent of the probe combinations. The signal strength of the lower-f"even" mode is larger with horizontally placed probes pointing to a prolate cavity deformation. A theoretical underpinning of the mode features are given in the next section.





ANALYSIS VIA MATHIEU FUNCTIONS

It can be to assume that the resonance splitting is caused by deforming a perfectly cylindrical tube/cavity into a prolate ellipse with the major axis, a, and the minor axis, b. The deformation is characterized by an eccentricity, $\varepsilon = \sqrt{1 - (b/a)^2}$ or in terms of the focal length, $h = \varepsilon a$. The deformation retains the perimeter, and is found as $P = 4aE(\varepsilon^2)$ with $E(\varepsilon^2)$ the complete elliptic integral of the second kind. Electromagnetic fields are solved in elliptic cylinder coordinates (ξ, η, z) related to Cartesian coordinates (x, y, z) by

$$x = h \cosh \xi \cos \eta , y = h \sinh \xi \sin \eta$$
 (1)

The surface $\xi = \text{constant}$ is a cylinder of elliptic cross section and $\eta = \text{constant}$ is a hyperbolic cylinder of two sheets with foci at $(\pm h, 0)$. The wall of the elliptic cylinder defines ξ_0 from $a = h \cosh \xi_0$ or $\varepsilon \cosh \xi_0 = 1$. The correlation of the elliptic cylinder with the cylindrical polar, (r, φ) , coordinates is complex but the simple relation of $\tan \varphi = \sqrt{1 - \varepsilon^2} \tan \eta$ is useful in the limit of small excentricity.

The field components for TE and TM time harmonic waves, are derived from scalar potentials written as the product of Mathieu functions [4],

 $u_{even} \propto Ce_m(q,\xi)ce_m(q,\eta), u_{odd} \propto Se_m(q,\xi)se_m(q,\eta)$ (2) with solutions of the coupled canonical and the modified (radial) Mathieu differential equations

$$\frac{\partial^2}{\partial \eta^2} \begin{cases} ce_m(\eta, q) \\ se_m(\mu, q) \end{cases} + \left(\begin{cases} a_m \\ b_m \end{cases} - 2q\cos 2\eta \right) \begin{cases} ce_m(\eta, q) \\ se_m(\eta, q) \end{cases} = 0 \quad (3)$$
$$\frac{\partial^2}{\partial \xi^2} \begin{cases} Ce_m(\xi, q) \\ Se_m(\xi, q) \end{cases} - \left(\begin{cases} a_m \\ b_m \end{cases} - 2q\cosh 2\xi \right) \begin{cases} Ce_m(\xi, q) \\ Se_m(\xi, q) \end{cases} = 0$$

where a_m and b_m are the even and odd characteristic values and

$$q = \frac{1}{4}h^2(k^2 - k_z^2), \ k = \omega/c \text{ and } k_z = 0 \text{ at cutoff.}$$
 (4)

The time harmonic electric field components, $e^{j\omega t}$, on the wall of the TE₁₁ mode are found as

$$E_{\xi} = \frac{Ce_{1}(\xi_{0}, q_{TE11})se_{1}'(\eta, q_{TE11})}{\sqrt{1 - (h/a)^{2}\cos^{2}\eta}}$$
$$E_{\xi} = \frac{C'e_{1}(\xi_{0}, q_{TE11})se_{1}(\eta, q_{TE11})}{\sqrt{1 - (h/a)^{2}\cos^{2}\eta}}$$
(5)

Assuming the use of electric probes for the S_{21} transmission measurements, the coupled TE₁₁ field components on the wall boundary of an elliptical structure are approximated by an even and an odd function,[5,6]

$$u_{e} \propto \frac{J_{1}(2\sqrt{q}\cosh\xi_{0})}{\sqrt{1-q^{2}\cos^{2}(\eta-\psi)}} \left[\cos(\eta-\psi) - (q/8)\cos(\eta-\psi) + ...\right]$$
$$u_{o} \propto \frac{J_{1}(2\sqrt{q}\cosh\xi_{0})}{\sqrt{1-q^{2}\cos^{2}(\eta-\psi)}} \left[\sin(\eta-\psi) - (q/8)\sin(\eta-\psi) + ...\right]$$

where $\eta \approx \varphi$ is the angular probe position, and ψ is the excitation angle versus the major axis in a prolate cavity. In first approximation, the ε -dependent *q*-parameter is given by the TE₁₁ cutoff frequency via $\varepsilon = 1/\cosh \xi_0$ as $2\sqrt{q} / \varepsilon = j'_{11}$, solution of the Bessel function, $J'_1(j'_{11}) = 0$ and $j'_{11} \approx 1.8412$, yielding the ratio of cutoff wavelength to the perfect tube radius as $\lambda_{co} / a \approx 3.4126$.

In contrast to the minimal effect of a small eccentricity on field shape, the cut-off frequencies are noticeably split into even and odd solutions with their ratio of wave length to radius defined in the literature as



Figure 4: Characteristic values of TE₁₁ mode, even dotted blue, odd dotted red, and fitted odd solid, respectively.

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The TE_{11} *q*-values have been obtained numerically by Kretschmar as [7]

$$q_{\text{CTE11}} \simeq 0.8476\varepsilon^2 - 0.0013\varepsilon^3 + 0.0379\varepsilon^4$$

 $q_{\text{STE11}} \simeq -0.0018\varepsilon + 0.8974\varepsilon^2 - 0.3679\varepsilon^3 + 0.1612\varepsilon^4$. (8) The interesting range of $\varepsilon \to 0$ is obtained by fitting the available numerical results in Eq. 8 to the circular tube with $\lim q_{811} / \varepsilon^2 = 0.8475$, yielding

$$q_{\text{STE11}}^{\diamond} \approx 0.8475\varepsilon^2 - 0.0888\varepsilon^3 - 0.2118\varepsilon^4$$
 (9)

Fig. 4 shows the characteristic values for the TE₁₁ mode over the interesting ε range. The odd q_{S11} (dotted red) and (solid) q_{S11}^{\diamond} from $0.05 \le \varepsilon \ge 0.5$ confirm a good fit.

For comparison, the wave propagation for TM_{11} follows from $j_{11} = 3.8317$ and has the $\lambda_{co} / a \approx 1.6398$ in the round wave guide. The characteristic values for the TM_{11} values were fitted (super script \diamond) over the available range in [7] yielding

$$q_{CTM11}^{\diamond} \simeq 3.6705\varepsilon^2 - 0.1140\varepsilon^3 + 1.4862\varepsilon^4 q_{STM11}^{\diamond} \simeq 3.6705\varepsilon^2 - 0.2084\varepsilon^3 + 3.8901\varepsilon^4$$
(10)

Frequency Splitting

Whereas the field shape is minimally affected by the ellipticity, the cutoff frequencies are split yielding the numerical expressions for their ratio [8]

$$\Delta (\lambda_{\rm CTE} / \lambda_{\rm STE}) \simeq 0.1032\varepsilon + 0.3054\varepsilon^2$$

$$\Delta (\lambda_{\rm CTM} / \lambda_{\rm STM}) \simeq 0.0129\varepsilon - 0.3268\varepsilon^2$$
(11)

The measured TE₁₁ frequency deviation at ~821 MHz of $\Delta \approx 6 \times 10^{-4}$ yield the numerical $\varepsilon = 0.0057$ and the focal distance $h = \varepsilon a \approx 0.3$ mm, similar to TESLA results [9].

CONCLUDING SIMULATION

The S_{21} transmission across the split 821 MHz dipole resonance is shown in Fig. 3 above. It led to the assumption that the intrinsic Q value is constant. Allowing in addition that the excitation of the modes with the coupling, κ , is done with equal probes at in and output and occurs in the same plane, defined by $\Upsilon = \eta - \psi$. The measured transmission can then be simulated according to Eq. 12,

$$S_{21} = \kappa \cos \Upsilon / \left\{ 1 + jQ \left(\frac{\omega}{\omega_{\rm e}} - \frac{\omega_{\rm e}}{\omega} \right) \right\} + \kappa \sin \Upsilon / \left\{ 1 + jQ \left(\frac{\omega}{\omega_{\rm o}} - \frac{\omega_{\rm o}}{\omega} \right) \right\}$$

with the measured split frequencies, $\omega_{\rm e}$, $\omega_{\rm o}$, while Q, κ and Υ are estimated or used as free fitting parameters. The highest Q TM₁₁ mode at ~1021 MHz is simulated in Mathematica with $Q \approx 43,000$ and its S_{21} is shown in Fig. 5 on top, confirming the validity of the model, excitation through equal κ to the split modes, although changing with the probe plane positions. The splitting is caused by the two mode polarizations with $\Upsilon \approx 10^{\circ}$ (blue) and 75° (red).



Figure 5: Simulated and measured $S_{21}(f)$ scans at 1021 MHz between cell 1 and 5 (blue) and cell 1 to PU (red).

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